RESEARCH

Advances in Difference Equations a SpringerOpen Journal

Open Access

Sampled-data state estimation for neural networks of neutral type

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Abstract

In this paper, the sampled-data state estimation is investigated for a class of neural networks of neutral type. By employing a suitable Lyapunov functional, a delay-dependent criterion is established to guarantee the existence of the sampled-data estimator. The estimator gain matrix can be obtained by solving linear matrix inequalities (LMIs). A numerical example is given to show the effectiveness of the proposed method.

Keywords: neural network; sampled-data; state estimation; neutral type; LMIs

1 Introduction

Over the past decades, considerable attention has been devoted to the study of artificial neural networks due to their extensive application in signal and image processing, pattern recognition, and combinatorial optimization [1, 2]. Numerous models of neural networks such as cellular neural networks, Hopfield-type neural networks, Cohen-Grossberg neural networks, and bidirectional associative memory neural networks have been extensively investigated in the literature [3–7].

In recent years, the state estimation problem has been a hot topic because of its great significance to estimation of the neuron states through available output measurements. Many results have been available. For example, the state estimation problem for delayed neural networks was discussed in [8]. Huang and Feng investigated the state estimation of recurrent neural networks with time-varying delay. A delay-dependent condition was derived to guarantee the existence of a desired state estimator [9, 10]. By constructing a Lyapunov-Krasovskii functional and using a convex combination technique, the exponential state estimation for Markovian jumping neural networks with time-varying discrete and distributed delays was investigated [11]. The state estimation of discrete-time neural networks and fuzzy neural networks was studied in [12–15]. Recently, the state estimation problem of neural networks of neutral type has motivated a great deal of interest. Park *et al.* investigated the state estimation of neural networks of neutral networks of neutral type with interval time-varying delays [16–18]. A delay-dependent condition was proposed for neural networks of neutral type with mixed time-varying delays and Markovian jumping parameters [19].

At the same time, the sampled-data state estimation of neural networks has gradually caused researchers' concern with the development of computer hardware technology. For instance, Hu investigated the sampled-data state estimation of delayed neural networks with Markovian jumping parameters [20], and in [21] the sampled-data state estimator was



©2014 Yang et al.; licensee Springer. This is an Open Access article distributed under the terms of the Creative Commons Attribution License (http://creativecommons.org/licenses/by/2.0), which permits unrestricted use, distribution, and reproduction in any medium, provided the original work is properly cited. designed for Markovian jumping fuzzy cellular neural networks with mode-dependent probabilistic time-varying delays. In order to effectively deal with the sampled-data, the author investigated the sampled-data state estimation of neural networks by using a discontinuous Lyapunov functional approach [22, 23]. It is worth mentioning that the approach to estimate the neuron states through the output sampled measurement needs less information from the network outputs, which can lead to a significant reduction of the information communication burden in the network. To the best of our knowledge, the sampled-data state estimation of neural networks of neutral type has not been investigated so far.

Motivated by the above discussion, in this paper, we aim to deal with the sampled-data state estimation problem for neural networks of neutral type. By utilizing Lyapunov functional, Jensen's inequalities and Schur complement, a delay-dependent criterion is given to guarantee the existence of the state estimator of neural networks of neutral type. The estimator gain matrix can be obtained in terms of LMIs.

The organization of the rest is as follows. In Section 2, the problem is formulated and some lemmas are introduced. In Section 3, some sufficient conditions are given to guarantee the existence of the sampled-data estimator. An example is given to exemplify the usefulness of our theoretical results in Section 4. And in the last section, Section 5, we give some conclusions.

2 Problem formulation

In this paper, the sampled-data state estimation problem will be investigated. Consider the following neural networks of neural type:

$$\dot{x}(t) = -Ax(t) + W_0 f(x(t)) + W_1 g(x(t-h(t))) + V\dot{x}(t-d(t)) + J,$$

$$y(t) = Cx(t),$$
(1)

where $x(t) = [x_1(t), ..., x_n(t)]^T \in \mathbb{R}^n$ is the neuron state vector associated with *n* neurons. $f(x(t)) = [f_1(x_1(t)), ..., f_n(x_n(t))]^T \in \mathbb{R}^n$ and $g(x(t-h(t))) = [g_1(x_1(t-h(t))), ..., g_n(x_n(t-h(t)))]^T \in \mathbb{R}^n$ are the neuron activation functions. $J = [J_1, ..., J_n]^T$ is the external input vector at time *t*. $y(t) \in \mathbb{R}^m$ is the measurement output. $A = \text{diag}(a_1, a_2, ..., a_n)$ is a positive diagonal matrix. W_0 , W_1 , and *V* are the interconnection matrices representing the weight coefficients of the neurons. $C \in \mathbb{R}^{m \times n}$ is a known constant matrix. h(t) > 0 and d(t) > 0 are axonal signal transmission delays that satisfy the following inequalities:

$$0 < h(t) \le \bar{h}, \qquad \dot{h}(t) \le h_D,$$

$$0 < d(t) \le \bar{d}, \qquad \dot{d}(t) \le d_D < 1$$

Assumption 1 The neuron activation functions $f(\cdot)$, $g(\cdot)$ satisfy the Lipschitz condition

$$\|f(x_1) - f(x_2)\| \le \|F(x_1 - x_2)\|,\$$

 $\|g(x_1) - g(x_2)\| \le \|G(x_1 - x_2)\|,\$

where $F \in \mathbb{R}^{n \times n}$ and $G \in \mathbb{R}^{n \times n}$ are the known constant matrices.

The aim of this paper is to present an efficient estimation algorithm to observe the neuron states from the available network sampling output. To this end, the following full order observer is proposed:

$$\dot{\hat{x}}(t) = -A\hat{x} + W_0 f\left(\hat{x}(t)\right) + W_1 g\left(\hat{x}\left(t - h(t)\right)\right) + V\dot{\hat{x}}\left(t - d(t)\right) + J + u(t),$$

$$\hat{y}(t) = C\hat{x}(t),$$
(2)

where $\hat{x}(t) \in \mathbb{R}^n$ is the estimation of neuron state x(t). $\hat{y}(t) \in \mathbb{R}^m$ is the estimated output vector. $u(t) \in \mathbb{R}^n$ is the control input, which is assumed to use sampled data as follows:

$$u(t) = K[y(t_k) - \hat{y}(t_k)]$$

= $KC[x(t_k) - \hat{x}(t_k)], \quad k = 0, 1, 2, ...,$ (3)

where t_k denotes the sampling instant and satisfies $\lim_{k\to\infty} t_k = \infty$.

Assumption 2 For $k \ge 0$, there exists a known positive constant $\overline{\tau}$ such that the sampling instant t_k satisfies $t_{k+1} - t_k \le \overline{\tau}$.

Define the error vector to be $e(t) = x(t) - \hat{x}(t)$, and

$$\begin{aligned} \phi(t) &= f\left(x(t)\right) - f\left(\hat{x}(t)\right), \\ \varphi(t) &= g\left(x(t)\right) - g\left(\hat{x}(t)\right). \end{aligned} \tag{4}$$

Let $\tau(t) = t - t_k$, $t_k \le t < t_{k+1}$, the controller (3) can be represented as follows:

$$u(t) = KCe(t_k)$$

= $KCe(t - \tau(t)), \quad t_k \le t < t_{k+1}.$ (5)

Therefore, the error dynamical system can be expressed by

$$\dot{e}(t) = -Ae(t) - KCe(t - \tau(t)) + V\dot{e}(t - d(t)) + W_0\phi(t) + W_1\phi(t - h(t)),$$
(6)

where $\tau(t)$ satisfies $0 \le \tau(t) \le \overline{\tau}$.

The following lemmas will be used in deriving the main results.

Lemma 1 ([19], Jensen's inequality) *For any constant matrix* $Q \in \mathbb{R}^{n \times n}$, $Q = Q^T > 0$, *scalar* b > 0, *and vector function* $x : [0, b] \to \mathbb{R}^n$, *one has*

$$-\int_0^b x^T(s)Qx(s)\,ds \leq -\frac{1}{b}\left[\int_0^b x(s)\,ds\right]^T Q\left[\int_0^b x(s)\,ds\right].$$

Lemma 2 ([20], Schur complement) *Given constant matrices* Ω_1 , Ω_2 , and Ω_3 , where $\Omega_1 = \Omega_1^T$ and $\Omega_2 > 0$, then

$$\Omega_1 + \Omega_3^T \Omega_2^{-1} \Omega_3 < 0 \tag{7}$$

if and only if

$$\begin{bmatrix} \Omega_1 & \Omega_3^T \\ \Omega_3 & -\Omega_2 \end{bmatrix} < 0.$$
(8)

3 Main results

In this section, we derive a new delay-dependent LMI criterion for the existence of state estimator (2) for neural networks (1).

Theorem 1 For given matrices F, G, positive scalars \overline{h} and $0 < \alpha < 1$, the error system (6) is globally asymptotically stable if there exist eight positive definite matrices P, Q_i (i = 1, 2, 3, 4, 5, 6, 7), two positive scalars β_1 and β_2 , and any matrix X satisfying the following LMI:

$$\begin{bmatrix} \Lambda_1 & \Lambda_3^T \\ * & -\Lambda_2 \end{bmatrix} < 0, \tag{9}$$

where

and then the estimator gain can be designed as $K = P^{-1}X$.

Proof Consider the following Lyapunov functional:

$$V = \sum_{i=1}^{8} V_i,$$
 (10)

where

$$V_{1} = e^{T}(t)Pe(t), \qquad V_{2} = \int_{t-\alpha h(t)}^{t} e^{T}(s)Q_{1}e(s) ds,$$

$$V_{3} = \int_{t-\bar{h}}^{t} \int_{s}^{t} \dot{e}^{T}(\theta)Q_{2}\dot{e}(\theta) d\theta ds, \qquad V_{4} = \int_{-\bar{\tau}}^{0} \int_{t+\theta}^{t} \dot{e}^{T}(s)Q_{3}\dot{e}(s) ds d\theta,$$

$$V_{5} = \int_{t-h(t)}^{t} \varphi^{T}(s)Q_{4}\varphi(s) ds, \qquad V_{6} = \int_{t-h(t)}^{t} e^{T}(s)Q_{5}e(s) ds,$$

$$V_{7} = \int_{t-\bar{h}}^{t} e^{T}(s)Q_{6}e(s) ds, \qquad V_{8} = \int_{t-d(t)}^{t} \dot{e}^{T}(s)Q_{7}\dot{e}(s) ds.$$

Calculating the derivative of V_i along the trajectories of system (6), we have

$$\dot{V}_{1} = 2e^{T}(t)P[-Ae(t) - KCe(t - \tau(t)) + V\dot{e}(t - d(t)) + W_{0}\phi(t) + W_{1}\phi(t - h(t))], \quad (11)$$
$$\dot{V}_{2} = e^{T}(t)Q_{1}e(t) - (1 - \alpha\dot{h}(t))e^{T}(t - \alpha h(t))Q_{1}e(t - \alpha h(t))$$

$$\leq e^{T}(t)Q_{1}e(t) - (1 - \alpha h_{D})e^{T}(t - \alpha h(t))Q_{1}e(t - \alpha h(t)), \qquad (12)$$

$$\dot{V}_3 = \bar{h}\dot{e}^T(t)Q_2\dot{e}(t) - \int_{t-\bar{h}}^t \dot{e}^T(s)Q_2\dot{e}(s)\,ds,\tag{13}$$

$$\dot{V}_4 = \bar{\tau} \dot{e}^T(t) Q_3 \dot{e}(t) - \int_{t-\bar{\tau}}^t \dot{e}^T(s) Q_3 \dot{e}(s) \, ds,\tag{14}$$

$$\dot{V}_{5} = \varphi^{T}(t)Q_{4}\varphi(t) - (1 - \dot{h}(t))\varphi^{T}(t - h(t))Q_{4}\varphi(t - h(t))$$

$$\leq e^{T}(t)G^{T}Q_{4}Ge(t) - (1 - h_{D})\varphi^{T}(t - h(t))Q_{4}\varphi(t - h(t)), \qquad (15)$$

$$V_{6} = e^{T}(t)Q_{5}e(t) - (1 - h(t))e^{T}(t - h(t))Q_{5}e(t - h(t))$$

$$\leq e^{T}(t)Q_{5}e(t) - (1 - h_{D})e^{T}(t - h(t))Q_{5}e(t - h(t)), \qquad (16)$$

$$\dot{V}_7 = e^T(t)Q_6e(t) - e^T(t - \bar{h})Q_6e(t - \bar{h}),$$
(17)

$$\dot{V}_{8} = \dot{e}^{T}(t)Q_{7}\dot{e}(t) - (1 - \dot{d}(t))\dot{e}^{T}(t - d(t))Q_{7}\dot{e}(t - d(t))$$

$$\leq \dot{e}^{T}(t)Q_{7}\dot{e}(t) - (1 - d_{D})\dot{e}^{T}(t - d(t))Q_{7}\dot{e}(t - d(t)).$$
(18)

Note that the Lyapunov functions $-\int_{t-\bar{h}}^{t} \dot{e}^{T}(s)Q_{2}\dot{e}(s) ds$ and $-\int_{t-\bar{\tau}}^{t} \dot{e}^{T}(s)Q_{3}\dot{e}(s) ds$ have the following relationship:

$$-\int_{t-\bar{h}}^{t} \dot{e}^{T}(s)Q_{2}\dot{e}(s) ds = -\int_{t-\alpha h(t)}^{t} \dot{e}^{T}(s)Q_{2}\dot{e}(s) ds - \int_{t-h(t)}^{t-\alpha h(t)} \dot{e}^{T}(s)Q_{2}\dot{e}(s) ds - \int_{t-\bar{h}}^{t-h(t)} \dot{e}^{T}(s)Q_{2}\dot{e}(s) ds,$$
(19)

$$-\int_{t-\bar{\tau}}^{t} \dot{e}^{T}(s)Q_{3}\dot{e}(s)\,ds = -\int_{t-\tau(t)}^{t} \dot{e}^{T}(s)Q_{3}\dot{e}(s)\,ds - \int_{t-\bar{\tau}}^{t-\tau(t)} \dot{e}^{T}(s)Q_{3}\dot{e}(s)\,ds.$$
(20)

By Lemma 1, we have

$$\begin{split} -\int_{t-\alpha h(t)}^{t} \dot{e}^{T}(s) Q_{2} \dot{e}(s) \, ds &\leq -(\alpha \bar{h})^{-1} \bigg[\int_{t-\alpha h(t)}^{t} \dot{e}(s) \, ds \bigg]^{T} Q_{2} \bigg[\int_{t-\alpha h(t)}^{t} \dot{e}(s) \, ds \bigg] \\ &\leq -(\alpha \bar{h})^{-1} [e(t) - e(t-\alpha h(t))]^{T} Q_{2} [e(t) - e(t-\alpha h(t))] \\ &= -(\alpha \bar{h})^{-1} [e^{T}(t) Q_{2} e(t) - 2e^{T}(t) Q_{2} e(t-\alpha h(t)) \\ &+ e^{T}(t-\alpha h(t)) Q_{2} e(t-\alpha h(t))], \\ -\int_{t-h(t)}^{t-\alpha h(t)} \dot{e}^{T}(s) Q_{2} \dot{e}(s) \, ds &\leq -(\bar{h}-\alpha \bar{h})^{-1} \bigg[\int_{t-h(t)}^{t-\alpha h(t)} \dot{e}(s) \, ds \bigg]^{T} Q_{2} \bigg[\int_{t-h(t)}^{t-\alpha h(t)} \dot{e}(s) \, ds \bigg] \\ &\leq -(\bar{h}-\alpha \bar{h})^{-1} [e(t-\alpha h(t)) - e(t-h(t))]^{T} \\ &\times Q_{2} [e(t-\alpha h(t)) - e(t-h(t))] \\ &= -(\bar{h}-\alpha \bar{h})^{-1} [e^{T}(t-\alpha h(t)) Q_{2} e(t-\alpha h(t)) \\ &- 2e^{T}(t-\alpha h(t)) Q_{2} e(t-h(t)) \bigg] \\ &= -(\bar{h}-\alpha \bar{h})^{-1} [e^{T}(t-\alpha h(t)) Q_{2} e(t-\alpha h(t)) \\ &- 2e^{T}(t-\alpha h(t)) Q_{2} e(t-h(t)) \bigg] , \\ -\int_{t-\bar{h}}^{t-h(t)} \dot{e}^{T}(s) Q_{2} \dot{e}(s) \, ds \leq -(\bar{h})^{-1} \bigg[\int_{t-\bar{h}}^{t-h(t)} \dot{e}(s) \, ds \bigg]^{T} Q_{2} \bigg[\int_{t-\bar{h}}^{t-h(t)} \dot{e}(s) \, ds \bigg] \\ &\leq -(\bar{h})^{-1} [e(t-h(t)) - e(t-\bar{h})]^{T} \\ &\times Q_{2} [e(t-h(t)) - e(t-\bar{h})]^{T} \\ &\times Q_{2} [e(t-h(t)) - e(t-\bar{h})] \\ &= -(\bar{h})^{-1} [e^{T}(t-h(t)) Q_{2} e(t-h(t)) \\ &- 2e^{T}(t-h(t)) Q_{2} e(t-\bar{h}) + e^{T}(t-\bar{h}) Q_{2} e(t-\bar{h})], \\ -\int_{t-\tau(t)}^{t} \dot{e}^{T}(s) Q_{3} \dot{e}(s) \, ds \leq -\frac{1}{\tau(t)} \bigg[\int_{t-\tau(t)}^{t-h(t)} \dot{e}(s) \, ds \bigg]^{T} Q_{3} \bigg[\int_{t-\tau(t)}^{t} \dot{e}(s) \, ds \bigg] \\ &\leq -\frac{1}{\tau} \bigg[e(t) - e(t-\tau(t)) \bigg]^{T} Q_{3} [e(t) - e(t-\tau(t))] \bigg] \\ &= -\frac{1}{\tau} \bigg[e^{T}(t) Q_{3} e(t) - 2e^{T}(t) Q_{3} e(t-\tau(t)) \\ &+ e^{T}(t-\tau(t)) Q_{3} e(t-\tau(t)) \bigg], \\ (22) \\ -\int_{t-\tau}^{t-\tau(t)} \dot{e}^{T}(s) Q_{3} \dot{e}(s) \, ds \leq -\frac{1}{\tau} \bigg[e(t-\tau(t)) - e(t-\bar{\tau}) \bigg]^{T} Q_{3} [e(t-\tau(t)) - e(t-\bar{\tau})] \bigg] \\ &= -\frac{1}{\tau} \bigg[e^{T}(t-\tau(t)) Q_{3} e(t-\tau(t)) - 2e^{T}(t-\tau(t)) Q_{3} e(t-\bar{\tau}) \\ &+ e^{T}(t-\tau(t)) Q_{3} e(t-\tau(t)) \bigg]. \end{aligned}$$

From Assumption 1 and (4), for positive scalars β_i (i = 1, 2), we have the following inequalities:

$$\beta_1 \left[e^T(t) F^T F e(t) - \phi^T(t) \phi(t) \right] \ge 0,$$

$$\beta_2 \left[e^T \left(t - h(t) \right) G^T G e(t - h(t)) - \varphi^T \left(t - h(t) \right) \varphi(t - h(t)) \right] \ge 0.$$
(23)

By combining (10)-(23), it is not difficult to deduce that

$$\begin{split} \dot{V} &= \sum_{i=1}^{8} \dot{V}_{i} \\ &\leq e^{T}(t) \bigg[-PA - A^{T}P + Q_{1} + Q_{5} + Q_{6} + G^{T}Q_{4}G - \frac{1}{\bar{\tau}}Q_{3} - (\alpha\bar{h})^{-1}Q_{2} \\ &+ \beta_{1}F^{T}F \bigg] e(t) + 2e^{T}(t) \bigg[-PKC + \frac{1}{\bar{\tau}}Q_{3} \bigg] e(t - \tau(t)) \\ &- e^{T}(t - \tau(t)) \bigg[\frac{2}{\bar{\tau}}Q_{3} \bigg] e(t - \tau(t)) + 2e^{T}(t) [PV] \dot{e}(t - d(t)) \\ &- \dot{e}^{T}(t - d(t)) \big[(1 - d_{D})Q_{7} \big] \dot{e}(t - d(t)) + 2e^{T}(t) [PW_{0}] \phi(t) \\ &- \phi^{T}(t) [\beta_{1}I] \phi(t) + 2e^{T}(t) [PW_{1}] \varphi(t - h(t)) \\ &- \varphi^{T}(t - h(t)) \big[(1 - h_{D})Q_{4} + \beta_{2}I \big] \varphi(t - h(t)) \\ &+ 2e^{T}(t) \big[(\alpha\bar{h})^{-1}Q_{2} \big] e(t - \alpha h(t)) - e^{T}(t - \alpha h(t)) \big[(1 - \alpha h_{D})Q_{1} \\ &+ (\bar{h} - \alpha\bar{h})^{-1}Q_{2} + (\alpha\bar{h})^{-1}Q_{2} \big] e(t - \alpha h(t)) \\ &+ 2e^{T}(t - \tau(t)) \bigg[\frac{1}{\bar{\tau}}Q_{3} \bigg] e(t - \bar{\tau}) - e^{T}(t - \bar{\tau}) \bigg[\frac{1}{\bar{\tau}}Q_{3} \bigg] e(t - \bar{\tau}) \\ &+ 2e^{T}(t - \alpha h(t)) \big[(\bar{h} - \alpha\bar{h})^{-1}Q_{2} \big] e(t - h(t)) \\ &- e^{T}(t - h(t)) \big[(1 - h_{D})Q_{5} + (\bar{h} - \alpha\bar{h})^{-1}Q_{2} + (\bar{h})^{-1}Q_{2} + \beta_{2}G^{T}G \big] e(t - h(t)) \\ &+ 2e^{T}(t - h(t)) \big[\bar{h}^{-1}Q_{2} \big] e(t - \bar{h}) - e^{T}(t - \bar{h}) \big[\bar{h}^{-1}Q_{2} + Q_{6} \big] e(t - \bar{h}) \\ &+ \dot{e}^{T}(t) \big[\bar{h}Q_{2} + \bar{\tau}Q_{3} + Q_{7} \big] \dot{e}(t). \end{split}$$

Let $\Upsilon = [-A - KC \ 0 \ 0 \ 0 \ 0 \ V \ W_0 \ W_1]$, and $\eta^T(t) = [e^T(t) \ e^T(t - \tau(t)) \ e^T(t - \bar{\tau}) \ e^T(t - h(t)) \ e^T(t - \bar{h})) \ e^T(t - h(t)) \ e^T(t - h(t))]$. From (6), we have $\dot{e}(t) = \Upsilon \eta(t)$. Therefore,

$$\dot{e}^{T}(t)[\bar{h}Q_{2}+\bar{\tau}Q_{3}+Q_{7}]\dot{e}(t)=\eta^{T}(t)[\Upsilon^{T}(\bar{h}Q_{2}+\bar{\tau}Q_{3}+Q_{7})\Upsilon]\eta(t).$$
(25)

Thus, changing the variable as X = PK, equation (24) can be rewritten as

$$\dot{V} \le \eta^T(t) \Big[\Lambda_1 + \Upsilon^T (\bar{h}Q_2 + \bar{\tau}Q_3 + Q_7) \Upsilon \Big] \eta(t),$$
(26)

where Λ_1 can be seen from Theorem 1.

In view of the fact $-PQ_2^{-1}P \le -(2P - Q_2)$, $-PQ_3^{-1}P \le -(2P - Q_3)$, $-PQ_7^{-1}P \le -(2P - Q_7)$. From (9), we can obtain the following inequality:

$$\begin{bmatrix} \Lambda_1 & \Lambda_3^T \\ * & -\Lambda_6 \end{bmatrix} < 0, \tag{27}$$

where

$$\Lambda_6 = \begin{bmatrix} \bar{h} P^T Q_2^{-1} P & 0 & 0 \\ * & \bar{\tau} P^T Q_3^{-1} P & 0 \\ * & * & P^T Q_7^{-1} P \end{bmatrix}.$$

$$\begin{bmatrix} \Lambda_1 & \Lambda_4^T \\ * & -\Lambda_5 \end{bmatrix} < 0, \tag{28}$$

where

By Lemma 2, inequality (28) is equivalent to the inequality $[\Lambda_1 + \Upsilon^T(\bar{h}Q_2 + \bar{\tau}Q_3 + Q_7)\Upsilon] < 0$. This implies that the error dynamic (6) is asymptotically stable. This completes the proof.

Remark 1 In the paper, the controller $u(t) = K[y(t_k) - \hat{y}(t_k)]$ is designed by the sampled data, which needs less information from the network outputs than the controller in [16, 17]. The sampled-data estimation approach can lead to a significant reduction of the information communication burden in the network and save more computing cost.

To this end, the sampled-data estimation problem has been solved and the desired estimator can be designed by Theorem 1. In the next section, the effectiveness of the developed sampled-data estimation approach for the neural network of neutral type is shown by a numerical example.

4 Simulation example

In this section, a numerical example with simulation results is employed to demonstrate the effectiveness of the proposed method.

Example 1 Consider the neural networks of neutral type with the following parameters:

$$A = \begin{bmatrix} 3 & 0 & 0 \\ 0 & 2 & 0 \\ 0 & 0 & 5 \end{bmatrix}, \quad W_0 = \begin{bmatrix} 0.3 & -0.1 & 0 \\ 0.1 & 0.4 & -0.2 \\ -0.2 & 0.1 & 0.2 \end{bmatrix}, \quad W_1 = \begin{bmatrix} 0.1 & 1 & 0.2 \\ -0.1 & 0.3 & -0.1 \\ -0.2 & 0.1 & 0.2 \end{bmatrix},$$
$$V = \begin{bmatrix} 0.1 & -0.05 & 0.05 \\ 0.05 & 0.1 & -0.05 \\ 0.05 & 0.1 & 0.1 \end{bmatrix}, \quad J = \begin{bmatrix} 2\sin(\pi t) + 0.03t^2 \\ -\sin(4t) - 0.04t^2 \\ 1.5\cos(4t) + 0.01t^2 \end{bmatrix}, \quad C = I,$$

$$f(x) = 0.5 \tanh(x), \qquad g(x(t - h(t))) = 0.4 \tanh(x(t - h(t))),$$

$$x(0) = [-1.5 \ 1 \ -1.5]^T, \qquad h(t) = 0.36(1 - \sin(t)),$$

$$d(t) = 0.25(1 - \sin(t)), \qquad \bar{\tau} = 0.05, \qquad \alpha = 0.5.$$

From the parameters above, it can be verified that F = 0.5I, G = 0.4I. By solving LMI given in Theorem 1, the feasible solutions are obtained as follows:

$$\begin{split} P &= \begin{bmatrix} 22.5358 & -1.1052 & 0.0140 \\ -1.1052 & 32.2373 & 0.6453 \\ 0.0140 & 0.6453 & 14.9831 \end{bmatrix}, \qquad Q_1 = \begin{bmatrix} 8.4845 & -0.0169 & -0.0747 \\ -0.0169 & 10.5790 & 0.1719 \\ -0.0747 & 0.1719 & 5.6045 \end{bmatrix}, \\ Q_2 &= \begin{bmatrix} 15.2407 & -0.6559 & -0.1345 \\ -0.6559 & 19.8115 & 0.2071 \\ -0.1345 & 0.2071 & 11.4000 \end{bmatrix}, \qquad Q_3 = \begin{bmatrix} 18.3501 & -0.1527 & 0.1320 \\ -0.1527 & 19.0875 & 0.0926 \\ 0.1320 & 0.0926 & 16.9542 \end{bmatrix}, \\ Q_4 &= \begin{bmatrix} 11.8617 & 0.2203 & 0.0955 \\ 0.2203 & 18.7959 & 1.1519 \\ 0.0955 & 1.1519 & 9.2649 \end{bmatrix}, \qquad Q_5 = \begin{bmatrix} 6.8482 & -0.0052 & -0.0725 \\ -0.0052 & 9.5406 & 0.1891 \\ -0.0725 & 0.1891 & 3.8875 \end{bmatrix}, \\ Q_6 &= \begin{bmatrix} 8.3328 & -0.0359 & -0.0743 \\ -0.0359 & 10.8433 & 0.1788 \\ -0.0743 & 0.1788 & 5.4375 \end{bmatrix}, \qquad Q_7 = \begin{bmatrix} 6.1224 & -0.3327 & 0.1054 \\ -0.3327 & 10.1217 & -0.1539 \\ 0.1054 & -0.1539 & 4.1545 \end{bmatrix}, \\ X &= \begin{bmatrix} -29.6833 & 0.2433 & -0.3916 \\ 0.2433 & -15.3925 & 0.9746 \\ -0.3916 & 0.9746 & -53.1328 \end{bmatrix}, \qquad \beta_1 = 18.9090, \qquad \beta_2 = 23.3269. \end{split}$$

Thus, the estimator gain matrix *K* is as follows:

$$K = P^{-1}X = \begin{bmatrix} -1.3190 & -0.0128 & -0.0102 \\ -0.0372 & -0.4796 & 0.1009 \\ -0.0233 & 0.0857 & -3.5505 \end{bmatrix}.$$

The simulation results are shown in Figures 1 and 2. It is easy to find that the responses of the state estimators track to true states very quickly.





5 Conclusion

In this paper, the sampled-data state estimation has been investigated for a class of neural networks of neutral type. By employing a Lyapunov functional, a delay-dependent condition is established to guarantee the existence of the desired state estimator. The feasible solution can be obtained easily by Matlab LMI toolbox. In the end, the numerical example is given to show the effectiveness of the proposed estimator. This work can be extended to the state estimation problem of general systems. This will be done in the near future.

Competing interests

The authors declare that they have no competing interests.

Authors' contributions

CY carried out the main results of this paper and drafted the manuscript. YY directed the study and helped to inspect the manuscript. All authors read and approved the final manuscript.

Acknowledgements

This work was jointly supported by the National Natural Science Foundation of China under Grant 60875036, the Foundation of Key Laboratory of Advanced Process Control for Light Industry (Jiangnan University), Ministry of Education, P.R. China the Fundamental Research Funds for the Central Universities (JUSRP51317B, JUDCF13042).

Received: 18 January 2014 Accepted: 21 April 2014 Published: 09 May 2014

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10.1186/1687-1847-2014-138

Cite this article as: Yang et al.: Sampled-data state estimation for neural networks of neutral type. Advances in Difference Equations 2014, 2014:138

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