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Passivity analysis and state estimation for a class of memristor-based neural networks with multiple proportional delays

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Abstract

This paper is concerned with the problem of a passivity analysis for a class of memristor-based neural networks with multiple proportional delays and the state estimator is designed for the memristive system through the available output measurements. By constructing a proper Lyapunov-Krasovskii functional, new criteria are obtained for the passivity and state estimation of the memristive neural networks. Finally, a numerical example is given to illustrate the feasibility of the theoretical results.

Keywords: memristor-based neural networks; passivity; state estimation; proportional delays

1 Introduction

In 1971, the memristor was proposed by Chua [1] as the fourth passive circuit element. Compared with the three conventional fundamental circuit elements (resistor, inductor, and capacitor), the memristor is a nonlinear time-varying element. The first practical memristor device was found by Strukov *et al.* [2] in 2008. The memristor presents the relationship between the charge (q) and flux (φ), *i.e.*, $d\varphi = M dq$. The memristor retains its most recent value when the voltage is turned off, so it re-expresses the retained value when it is turned on. This feature makes them useful as energy-saving devices that can compete with flash memory and other static memory devices. Some classes of memristors also have nonlinear response characteristics which makes them doubly suitable as artificial neurons.

Existing results show that the electronic synapses and neurons can represent important functionalities of their biological counterparts [3]. Recently, the simulation of different kinds of memristors has developed rapidly and the studies of memristive neural networks have aroused more attention [3–15]. In [16], Wang *et al.* considered the following memristive neurodynamic system:

$$\dot{x}_{i}(t) = -\frac{1}{\mathbb{C}_{i}} \left[\sum_{j=1}^{n} \frac{1}{\mathbb{R}_{ij}} \times \operatorname{sign}_{ij} + \frac{1}{\mathcal{R}_{i}} \right] x_{i}(t) + \frac{1}{\mathbb{C}_{i}} \sum_{j=1}^{n} \frac{g_{j}(x_{j}(t))}{\mathbb{R}_{ij}} \times \operatorname{sign}_{ij},$$

$$t \ge 0, i = 1, 2, \dots, n,$$

$$(1.1)$$



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where $x_i(t)$ is the voltage of the capacitor \mathbb{C}_i at time t, \mathbb{R}_{ij} denotes the resistor through the feedback function $g_i(x_i(t))$ and $x_i(t)$, \mathcal{R}_i represents the parallel-resistor corresponding the capacitor \mathbb{C}_i , and

$$\operatorname{sign}_{ij} = \begin{cases} 1, & i \neq j, \\ -1, & i = j. \end{cases}$$

Based on two kinds of memductance functions, finite-time stability criteria were obtained for memristive neural networks with stochastic perturbations. The analysis employed differential inclusions theory, finite-time stability theorem, linear matrix inequalities, and the Lyapunov functional method.

In practice, due to the finite speed of information processing and the inherent communication time of neurons, time delays are frequently encountered in many biological and artificial neural networks. As correctly pointed out in [17–20], time delays may cause undesirable dynamical network behaviors such as oscillations, divergences, and so on. Consequently, it is valuable to investigate the problems of dynamics analysis for neural networks with time delay. By using different approaches, many significant results have been reported such as the global robust passivity analysis in [17] and the passivity analysis in [18–23]. In order to investigate the effect of time delay upon the dynamics of the memristive neural networks, Wu and Zeng [10] considered the following system:

$$\dot{x}_{i}(t) = -x_{i}(t) + \sum_{j=1}^{n} a_{ij}(x_{i}(t))f_{j}(x_{j}(t)) + \sum_{j=1}^{n} b_{ij}(x_{i}(t))f_{j}(x_{j}(t-\tau_{j})) + u_{i}(t),$$

$$z_{i}(t) = f_{i}(x_{i}(t)) + f_{i}(x_{i}(t-\tau_{i})) + u_{i}(t), \quad t \ge 0, i = 1, 2, ..., n,$$
(1.2)

where τ_j is the time delay that satisfies $0 \le \tau_j \le \tau$ ($\tau \ge 0$ is a constant). $u_i(t)$ and $z_i(t)$ denote external input and output, respectively, $f_j(\cdot)$ is the neuron activation function satisfying $f_j(0) = 0$, $a_{ij}(x_i(t))$ and $b_{ij}(x_i(t))$ represent memristor-based weights. Based on the theories of nonsmooth analysis and linear matrix inequalities, by using a suitable Lyapunov functional, the exponential passivity was studied for the memristive neural networks with time delays.

Proportional delay is a time-varying delay with time proportionality, which is usually required in web quality of service routing decisions. Different from the constant time delay, bounded time-varying delay, and distributed delay, the proportional delay is unbounded and time-varying. Recently, routing algorithms were obtained in [24, 25] based on neural networks. These routing algorithms were proven to be able to get the exact solutions for the problems which have a high parallelism. As well known, there exists a spatial extent in neural networks as the presence of parallel pathways of different axonal sizes and lengths. Accordingly, it is reasonable to use continuously proportional delays to describe the topology structure and parameters of neural networks. The proportional delayed systems have aroused many scholars' attention [7, 26, 27]. In [7], Wang *et al.* investigated the following memristive neural networks with multiple proportional delays:

$$\dot{x}_{i}(t) = -x_{i}(t) + \sum_{j=1}^{n} a_{ij}(x_{i}(t))f_{j}(x_{j}(t)) + \sum_{j=1}^{n} b_{ij}(x_{i}(t))f_{j}(x_{j}(q_{j}t)), \quad i = 1, 2, ..., n,$$
(1.3)

where

$$a_{ij}(x_i(t)) = \frac{\mathbb{M}_{ij}}{C_i} \times \operatorname{sign}_{ij}, \qquad b_{ij}(x_i(t)) = \frac{\mathbb{W}_{ij}}{C_i} \times \operatorname{sign}_{ij}, \qquad \operatorname{sign}_{ij} = \begin{cases} 1, & i \neq j, \\ -1, & i = j, \end{cases}$$

 $x_i(t)$ is the voltage of the capacitor C_i at time t, \mathbb{M}_{ij} and \mathbb{W}_{ij} denote the memductances of memristor R_{ij} and \hat{R}_{ij} , respectively. R_{ij} represents the memristor between the neuron activation function $f_j(x_j(t))$ and $x_i(t)$. \hat{R}_{ij} represents the memristor between the neuron activation functions $f_j(x_j(q_jt))$ and $x_i(t)$. q_jt is a proportional delay factor satisfying $q_jt = t - (1 - q_j)t$, in which $(1 - q_j)t$ corresponds to the time delay required in processing and transmitting a signal from the *j*th neuron to the *i*th neuron. By using the differential inclusion theory to handle the memristive neural networks with discontinuous right-hand side, several criteria ensuring anti-synchronization of memristive neural networks with multiple proportional delays were presented.

As well known, memristor-based neural networks are state-dependent switched nonlinear systems. The passivity theory, originating from circuit theory, is the most important issue in the analysis and design of switched systems. In the passivity theory, the passivity means that systems can keep internally stable. Therefore, the passivity theory provides a tool to analyze the stability of control systems and it has been applied in many areas. Based on the passivity theory, the authors in [28] dealt with the problems of sliding mode control for uncertain singular time-delay systems. In [29], the authors designed a modedependent state feedback controller by applying the passivity theory. The design of a passive controller based on the passivity analysis of nonlinear systems has become an effective way to solve practical engineering problems, for example, passivity-based control of threephase AC/DC voltage-source converters. For details, the reader is referred to [30, 31] and the references therein. As state-dependent switched nonlinear systems, memristive neural networks include too many subsystems. The product of input and output is utilized as the energy provision of the passive systems, which embodies the energy attenuation character. Passivity analysis of memristive neural networks provides a way to understand the complex brain functionalities with the adoption of memristor-MOS technology designs [32]. In recent years, although some research results have been concerned with the passivity problem for memristive neural networks, little attention has been paid to the problem of state estimation for memristor-based neural networks with multiple proportional delays based on passivity theory. Due to the importance of proportional delays in web quality of service routing decisions, we carry out this work to shorten the gap.

Motivated by the works in [10, 32, 33] and the circuit design in [16, 34–36], a class of memristive neural networks with multiple proportional delays is considered. By constructing a proper Lyapunov-Krasovskii functional, passivity criteria are obtained for the given delayed system, and then the state estimator is designed for the memristive system through available output measurements based on passivity theory.

The rest of this paper is organized as follows. In Section 2, the corresponding delayed neurodynamic equation for the presented memristive circuit is established and preliminaries are given. The theoretic results are derived in Section 3. In Section 4, the validity of the theoretical analysis is discussed through a numerical example.

2 Model description and preliminaries

In this paper, we consider the following memristor-based neural networks with multiple proportional delays which is unbounded:

$$\dot{x}_{i}(t) = -d_{i}x_{i}(t) + \sum_{j=1}^{n} a_{ij}(x_{i}(t))f_{j}(x_{j}(t)) + \sum_{j=1}^{n} b_{ij}(x_{i}(t))f_{j}(x_{j}(\rho_{j}(t))) + I_{i}(t),$$

$$y_{i}(t) = f_{i}(x_{i}(t)), \quad t \ge 0, i = 1, 2, ..., n,$$
(2.1)

or equivalently

$$\dot{x}(t) = -Dx(t) + A(x(t))f(x(t)) + B(x(t))f(x(\rho(t))) + I(t),$$

$$y(t) = f(x(t)), \quad t \ge 0, i = 1, 2, \dots, n.$$
(2.2)

System (2.1) can be implemented by the large-scale integration circuits as shown in Figure 1, and system (2.1) can be obtained using Kirchhoff's current law. $x(t) = [x_1(t), x_2(t), ..., x_n(t)]^T \in \mathbb{R}^n$ is the voltage of the capacitor C_i at time t, $\rho(t) = [\rho_1(t), \rho_2(t), ..., \rho_n(t)]^T$ represents the proportionality delay factor and $0 < q_j \le 1$, $\rho_j(t) = t - (1 - q_j)t$, where $(1 - q_j)t$ is the time delay required in processing and transmitting a signal between neurons and $(1 - q_j)t$ is a time-varying continuous function satisfying $(1 - q_j)t \rightarrow +\infty$ as $q_j \neq 1$, $t \rightarrow +\infty$. $f_j(x_j(t))$ and $f_j(x_j(\rho_j(t)))$ are the nonlinear neuron activation functions of the *j*th neuron at time *t* and q_jt , respectively. $I(t) = [I_1(t), I_2(t), ..., I_n(t)]^T$ is the input vector at time *t*, $y(t) = [y_1(t), y_2(t), ..., y_n(t)]^T \in \mathbb{R}^n$ denotes the output vector of the networks. $D = \text{diag}(d_1, d_2, ..., d_n)$ describes the rate with which each neuron will reset its potential to the resting state in isolation when disconnected from the networks and external inputs. $A(x(t)) = (a_{ij}(x_i(t)))_{n \times n}$ and $B(x(t)) = (a_{ij}(x_i(t)))_{n \times n}$ represent the memristor-based connec-



tion weight matrices, and

$$a_{ij}(x_i(t)) = \frac{\mathbf{W}_{ij}}{\mathbf{C}_i} \operatorname{sign}_{ij}, \qquad b_{ij}(x_i(t)) = \frac{\mathbf{M}_{ij}}{\mathbf{C}_i} \operatorname{sign}_{ij}, \qquad \operatorname{sign}_{ij} = \begin{cases} 1, & i \neq j, \\ -1, & i = j, \end{cases}$$
(2.3)

in which \mathbf{W}_{ij} and \mathbf{M}_{ij} denote the memductances of memristors \mathbf{R}_{ij} and \mathbf{F}_{ij} . Here \mathbf{R}_{ij} represents the memristor across $x_j(t)$ and the feedback function $f_j(x_j(t))$, and \mathbf{F}_{ij} represents the memristor across $x_j(t)$ and the feedback function $f_j(x_j(\rho_j(t)))$. According to the feature of the memristor and the current voltage characteristic, we have

$$a_{ij}(x_{i}(t)) = \begin{cases} \hat{a}_{ij}, \operatorname{sign} \frac{df_{j}(x_{j}(t))}{dt} - \frac{dx_{i}(t)}{dt} \leq 0, \\ \check{a}_{ij}, \operatorname{sign} \frac{df_{j}(x_{l}(t))}{dt} - \frac{dx_{i}(t)}{dt} > 0, \end{cases}$$

$$b_{ij}(x_{i}(t)) = \begin{cases} \hat{b}_{ij}, \operatorname{sign} \frac{df_{j}(x_{j}(\rho_{j}(t)))}{dt} - \frac{dx_{i}(t)}{dt} \leq 0, \\ \check{b}_{ij}, \operatorname{sign} \frac{df_{j}(x_{j}(\rho_{j}(t)))}{dt} - \frac{dx_{i}(t)}{dt} > 0 \end{cases}$$
(2.4)

for i, j = 1, 2, ..., n, where $\hat{a}_{ij}, \check{a}_{ij}, \hat{b}_{ij}$ and \check{b}_{ij} are constants. Obviously, the memristive neural network model (2.1) is a state-dependent switched nonlinear system.

System (2.1) is established based on the following assumption:

(A1): The neural activation function $f_i(\cdot)$ satisfies Lipschitz condition with a Lipschitz constant k_i , *i.e.*,

$$|f_i(u) - f_i(v)| \le k_i |u - v|, \quad i = 1, 2, \dots, n, \forall u, v \in R \text{ and } u \ne v,$$

where *u* and *v* are known constant scalars and $f_i(0) = 0$.

Denote $D = \text{diag}(d_1, d_2, ..., d_n)$. I_n is an $n \times n$ identity matrix. For the symmetric matrix **T**, **T** > 0 (**T** < 0) means that **T** is a positive definite (negative definite) matrix, and **T** \geq 0 (**T** \leq 0) means that **T** is a semi-positive definite (semi-negative definite) matrix. For matrices **Q** = $(q_{ij})_{n \times n}$ and **H** = $(h_{ij})_{n \times n}$, **Q** \gg **H** (**Q** \ll **H**) means that $q_{ij} \geq h_{ij}$ ($q_{ij} \leq h_{ij}$), for i, j = 1, 2, ..., n. And by the interval matrix [**Q**, **H**], it follows that **Q** \ll **H**. For any matrix **L** = $(l_{ij})_{n \times n} \in [\mathbf{Q}, \mathbf{H}]$, it means **Q** \ll **L** \ll **H**, *i.e.*, $q_{ij} \leq l_{ij} \leq h_{ij}$ for i, j = 1, 2, ..., n. The symmetric terms in a symmetric matrix are denoted by '*.

Definition 1 ([37]) Let $G \subseteq \mathbb{R}^n$, $x \mapsto H(x)$ is called a set-valued map from $G \hookrightarrow \mathbb{R}^n$, if, to each point x of a set $G \subseteq \mathbb{R}^n$, there corresponds a nonempty set $H(x) \subseteq \mathbb{R}^n$.

Definition 2 ([37]) A set-valued map H with nonempty values is said to be upper semicontinuous at $x_0 \in G \subseteq \mathbb{R}^n$, if for any open set M containing $H(x_0)$, there exists a neighborhood N of x_0 such that $H(N) \subseteq M$. H(x) is said to have a closed (convex, compact) image if, for each $x \in G$, H(x) is closed (convex, compact).

Definition 3 ([38]) For the differential system $\dot{x} = h(t, x)$, where h(t, x) is discontinuous in x, the set-valued map of h(t, x) is defined as

$$H(t,x) = \bigcap_{\epsilon>0} \bigcap_{\mu(M)=0} \operatorname{co}[h(B(x,\epsilon) \setminus M)],$$

where $B(x, \epsilon) = \{y : ||y - x|| \le \epsilon\}$ is the ball of center *x* and radius ϵ ; the intersection is taken over all sets *M* of measure zero and over all $\epsilon > 0$; and $\mu(M)$ is the Lebesgue measure of the set *M*.

A Filippov solution of system $\dot{x} = h(t, x)$, with initial condition $x(0) = x_0$ is absolutely continuous on any subinterval $t \in [t_1, t_2]$ of [0, T], which satisfies $x(0) = x_0$, and the differential inclusion:

$$\dot{x} \in H(t, x)$$
 for a.a. $t \in [0, T]$.

Definition 4 A constant vector $x^* = (x_1^*, x_2^*, ..., x_n^*)$ is said to be an equilibrium point of system (2.1), if for i = 1, 2, ..., n,

$$0 \in -d_i(x_i^*) + \sum_{j=1}^n K(a_{ij}(x_i^*))f_j(x_j^*) + \sum_{j=1}^n K(b_{ij}(x_i^*))f_j(x_j^*) + I_i(t),$$

where $K(\mathcal{P})$ denotes the closure of the convex hull of the set \mathcal{P} .

Lemma 1 ([39]) System (2.1) has at least one equilibrium point.

Proof Consider the following set-valued map:

$$x_{i} \multimap H_{i}(t, x_{i}) := \frac{1}{d_{i}(x_{i}^{*})} \left[\sum_{j=1}^{n} K(a_{ij}(x_{i}^{*})) f_{j}(x_{j}^{*}) + \sum_{j=1}^{n} K(b_{ij}(x_{i}^{*})) f_{j}(x_{j}^{*}) + I_{i}(t) \right],$$

$$i = 1, 2, \dots, n.$$

From Definitions 1-3, we can see that the set-valued map $x_i \multimap H_i(t, x_i)$ is upper semicontinuous with nonempty compact convex values, and

$$H(t,x) = (H_1(t,x_1), H_2(t,x_2), \dots, H_n(t,x_n))^T$$

is bounded since $f_j(\cdot)$ is bounded. According to Kakutani's fixed point theorem, there exists at least one fixed point of H(t, x), which is also an equilibrium point of system (2.1).

Considering the following nonlinear transformation:

$$z_i(t) = x_i(e^t), \tag{2.5}$$

we can prove that system (2.1) is equivalent to the following neural networks with multiple constant delays and time-varying coefficients:

$$\dot{z}_{i}(t) = e^{t} \left[-d_{i}z_{i}(t) + \sum_{j=1}^{n} a_{ij}(z_{i}(t))f_{j}(z_{j}(t)) + \sum_{j=1}^{n} b_{ij}(z_{i}(t))f_{j}(z_{j}(t-\tau_{j})) + u_{i}(t) \right],$$

$$y_{i}(t) = f_{i}(z_{i}(t)), \quad t \ge 0, i = 1, 2, ..., n,$$
(2.6)

in which $\tau_i = -\log q_i \ge 0$, $u_i(t) = I_i(e^t)$.

The initial conditions of system (2.6) are assumed to be

$$z_i(t) = \phi_i(t), \quad t \in [-\tau, 0], \tau = \max_{0 \le i \le n} \{\tau_j\}, i, j = 1, 2, \dots, n,$$

where $\phi_i(t) = (\phi_1(t), \phi_2(t), \dots, \phi_n(t)) \in C([-\tau, 0], \mathbb{R}^n)$, \mathbb{R}^n is *n*-dimensional Euclidean space. $C([-\rho, 0], \mathbb{R}^n)$ is a Banach space of all continuous functions with the Euclidean norm.

It is easy to see that the weight $a_{ij}(x_i(t))$ switches between two different constant values \hat{a}_{ij} and \check{a}_{ij} , and $b_{ij}(x_i(t))$ switches between \hat{b}_{ij} and \check{b}_{ij} . Consequently, the combination number of the possible form of the connection weight matrices A(x(t)) and B(x(t)) is 2^{2n^2} . So, we can order these 2^{2n^2} cases in the following way:

$$(A_1, B_1),$$
 $(A_2, B_2),$..., $(A_{2^{2n^2}}, B_{2^{2n^2}}).$

Then, at any fixed time $t \ge 0$, the form of A(x(t)) and B(x(t)) must be one of the 2^{2n^2} cases. For $l \in \{1, 2, ..., 2^{2n^2}\}$, at any fixed time t, we define the characteristic function of A_l and B_l as follows:

$$\pi_{l}(t) = \begin{cases} 1, & A(x(t)) = A_{l}, & B(x(t)) = B_{l}, \\ 0, & \text{otherwise.} \end{cases}$$
(2.7)

We can easily obtain $\sum_{l=1}^{2^{2n^2}} \pi_l(t) = 1$. Thus, the memristive neural network model (2.2) can be rewritten as

$$\dot{x}(t) = \sum_{l=1}^{2^{2n^2}} \pi_l(t) \Big[-Dx(t) + A_l f(x(t)) + B_l f(x(qt)) + I(t) \Big],$$

$$y(t) = f(x(t)), \quad t \ge 0, i = 1, 2, ..., n.$$
(2.8)

According to (2.2) and (2.8), we can obtain

$$A(x(t)) = \sum_{l=1}^{2^{2n^2}} \pi_l(t) A_l, \qquad B(x(t)) = \sum_{l=1}^{2^{2n^2}} \pi_l(t) B_l.$$
(2.9)

Definition 5 ([40]) The system (2.6) is said to be passive if there exists a scalar $\gamma \ge 0$ such that, for all $t_p \ge 0$ and all solutions of system (2.6) with z(0) = 0, the formula

$$2\int_0^{t_p} y^T(s)u(s)\,\mathrm{d}s \ge -\gamma \int_0^{t_p} u^T(s)u(s)\,\mathrm{d}s$$

holds, where $y(t) = (y_1(t), y_2(t), \dots, y_n(t))^T$, $u(t) = (u_1(t), u_2(t), \dots, u_n(t))^T$.

The product of input and output is regarded as the energy provision for the passivity of the systems in Definition 5, which embodies the energy attenuation character of system. By the control theories, we know that the passive properties of systems can keep the systems internally stable. Passive systems only burn energy while there is no energy production. Accordingly, by nature, passivity embodies a characteristic of the energy consumption of

system. We know that the power flow is usually made to meet the energy conservation and the passive systems do not produce energy [32], *i.e.*,

$$E_{input} + E_{initial} = E_{residual} + E_{dissipated}$$
.

3 Main results

In this section, we present our passivity analysis results of system (2.1) and develop a passivity-based approach dealing with the state estimation problem for system (2.1).

Denote by *I* an *n*-dimensional identity matrix. For convenience, we use the following notations:

$$\Gamma = \operatorname{diag}\left(\frac{d_1}{k_1}, \frac{d_2}{k_2}, \dots, \frac{d_n}{k_n}\right), \qquad \Lambda = \operatorname{diag}(\lambda_1, \lambda_2, \dots, \lambda_n), \qquad K = \operatorname{diag}(k_1, k_2, \dots, k_n),$$

where

$$\lambda_i = \begin{cases} 1, & i = j, \\ 0, & i \neq j. \end{cases}$$

3.1 Passivity analysis

Theorem 3.1 Suppose the assumption (A1) holds and

$$\Theta = \begin{pmatrix} \Pi & B_l \\ * & -I \end{pmatrix} < 0, \tag{3.1}$$

where $\Pi = A_l + A_l^T + \Lambda - \Lambda \Gamma - \Gamma^T \Lambda^T$, system (2.1) is passive in the sense of Definition 5.

Proof Define a Lyapunov-Krasovskii functional V(z(t)):

$$V(z(t)) = 2\sum_{i=1}^{n} e^{-t} \int_{0}^{z_{i}(t)} f_{i}(s) \,\mathrm{d}s + \sum_{i=1}^{n} \int_{t-\tau_{i}}^{t} f_{i}^{2}(z_{i}(s)) \,\mathrm{d}s.$$
(3.2)

Calculating the derivative of V(z(t)) along the positive half trajectory of system (2.6), we obtain

$$\begin{split} \dot{V}(z(t)) &- 2y^{T}(t)u(t) - \gamma u^{T}(t)u(t) \\ &= -2\sum_{i=1}^{n} e^{-t} \int_{0}^{z_{i}(t)} f_{i}(s) ds + 2\sum_{i=1}^{n} e^{-t} f_{i}(z_{i}(t)) \dot{z}_{i}(t) + \sum_{i=1}^{n} f_{i}^{2}(z_{i}(t)) - \sum_{i=1}^{n} f_{i}^{2}(z_{i}(t-\tau_{i})) \\ &- 2y^{T}(t)u(t) - \gamma u^{T}(t)u(t) \\ &\leq 2\sum_{i=1}^{n} e^{-t} f_{i}(z_{i}(t)) \dot{z}_{i}(t) + \sum_{i=1}^{n} f_{i}^{2}(z_{i}(t)) - \sum_{i=1}^{n} f_{i}^{2}(z_{i}(t-\tau_{i})) - 2y^{T}(t)u(t) - \gamma u^{T}(t)u(t) \\ &= 2\sum_{i=1}^{n} f_{i}(z_{i}(t)) \bigg[-d_{i}z_{i}(t) + \sum_{j=1}^{n} a_{ij}(z_{i}(t)) f_{j}(z_{j}(t)) + \sum_{j=1}^{n} b_{ij}(z_{i}(t)) f_{j}(z_{j}(t-\tau_{j})) + u_{i}(t) \bigg] \\ &+ \sum_{i=1}^{n} f_{i}^{2}(z_{i}(t)) - \sum_{i=1}^{n} f_{i}^{2}(z_{i}(t-\tau_{i})) - 2y^{T}(t)u(t) - \gamma u^{T}(t)u(t) \end{split}$$

(3.3)

$$= 2 \sum_{i=1}^{n} \left[-d_{i}z_{i}(t)f_{i}(z_{i}(t)) + \sum_{j=1}^{n} a_{ij}(z_{i}(t))f_{i}(z_{i}(t))f_{j}(z_{j}(t)) \right] \\ + \sum_{j=1}^{n} b_{ij}(z_{i}(t))f_{i}(z_{i}(t))f_{j}(z_{j}(t-\tau_{j})) \right] \\ + \sum_{i=1}^{n} f_{i}^{2}(z_{i}(t)) - \sum_{i=1}^{n} f_{i}^{2}(z_{i}(t-\tau_{i})) - \gamma u^{T}(t)u(t) \\ \leq 2 \sum_{i=1}^{n} \left[\left(\frac{1}{2} - \frac{d_{i}}{k_{i}} \right) f_{i}^{2}(z_{i}(t)) + \sum_{j=1}^{n} a_{ij}(z_{i}(t)) f_{i}(z_{i}(t)) f_{j}(z_{j}(t)) \right. \\ + \sum_{j=1}^{n} b_{ij}(z_{i}(t)) f_{i}(z_{i}(t)) f_{j}(z_{j}(t-\tau_{j})) \right] \\ - \sum_{i=1}^{n} f_{i}^{2}(z_{i}(t-\tau_{i})) - \gamma u^{T}(t)u(t) \\ = 2 \sum_{i=1}^{n} \left[\sum_{j=1}^{n} \left(a_{ij}(z_{i}(t)) + \lambda_{ij}\left(\frac{1}{2} - \frac{d_{i}}{k_{i}} \right) \right) f_{i}(z_{i}(t)) f_{j}(z_{j}(t)) \\ + \sum_{j=1}^{n} b_{ij}(z_{i}(t)) f_{i}(z_{i}(t)) f_{j}(z_{j}(t-\tau_{j})) \right] \\ - \sum_{i=1}^{n} f_{i}^{2}(z_{i}(t-\tau_{i})) - \gamma u^{T}(t)u(t) \\ = \eta^{T}(t) \widetilde{\Xi} \eta(t),$$

where

$$\eta(t) = \left[f(z(t)), f(z(t-\tau)), u(t)\right]^{T}, \qquad \widetilde{\Xi} = \begin{pmatrix} \widetilde{\Pi} & B(z(t)) & 0\\ * & -I & 0\\ * & * & -\gamma \end{pmatrix},$$

in which $\widetilde{\Pi} = A(z(t)) + A^T(z(t)) + \Lambda - \Lambda \Gamma - \Gamma^T \Lambda^T$. According to (2.9), we obtain

$$\widetilde{\Xi} = \sum_{l=1}^{2^{2n^2}} \pi_l(t) \begin{pmatrix} \Theta & 0 \\ * & -\gamma \end{pmatrix} < 0.$$
(3.4)

It follows from (3.3) and (3.4) that

$$\dot{V}(z(t)) - 2\gamma^{T}(t)u(t) - \gamma u^{T}(t)u(t) \le 0.$$
(3.5)

Integrating (3.5) with respect to t over the time period from 0 to $t_p,$ we have

$$2\int_{0}^{t_{p}} \gamma^{T}(s)u(s) \, \mathrm{d}s \geq V(t_{p}, z(t_{p})) - V(0, z(0)) - \gamma \int_{0}^{t_{p}} u^{T}(t)u(t) \, \mathrm{d}s$$

for z(0) = 0. Since V(0, z(0)) = 0 and $V(t_p, z(t_p)) \ge 0$, we get

$$2\int_0^{t_p} \gamma^T(s)u(s)\,\mathrm{d}s \geq -\gamma \int_0^{t_p} u^T(t)u(t)\,\mathrm{d}s.$$

Therefore, the memristive neural network (2.1) is passive in the sense of Definition 5. This completes the proof. $\hfill \Box$

Remark 1 In Theorem 3.1, Lyapunov stability theory is employed to investigate the passivity of system (2.1). Actually, we can find that passivity is a higher abstraction level of stability in the process of the proof, and the corresponding Lyapunov function (3.2) can be regarded as the storage function.

When the input is $I(t) = (I_1(t), I_2(t), \dots, I_n(t))^T = (0, 0, \dots, 0)^T$, by using standard arguments as Theorem 3.1, we can derive the stability conditions for system (2.1).

Corollary 3.1 Suppose the assumption (A1) holds and

$$\Theta = \begin{pmatrix} \Pi & B_l \\ * & -I \end{pmatrix} < 0, \tag{3.6}$$

where $\Pi = A_l + A_l^T + \Lambda - \Lambda \Gamma - \Gamma^T \Lambda^T$, system (2.1) is globally asymptotically stable while the input $I(t) = (I_1(t), I_2(t), \dots, I_n(t))^T = (0, 0, \dots, 0)^T$.

3.2 State estimation based on passivity

In the following, we design a state estimator for neural networks (2.1) based on passivity theory as follows:

$$\begin{aligned} \dot{\bar{x}}_{i}(t) &= -d_{i}\overline{x}_{i}(t) + \sum_{j=1}^{n} a_{ij}(\bar{x}_{i}(t))f_{j}(\bar{x}_{j}(t)) + \sum_{j=1}^{n} b_{ij}(\bar{x}_{i}(t))f_{j}(\bar{x}_{j}(\rho_{j}(t))) + I_{i}(t) \\ &+ m_{i}(y_{i}(t) - \bar{y}_{i}(t)) - J_{i}(t), \\ \bar{y}_{i}(t) &= f_{i}(\bar{x}_{i}(t)), \quad t \geq 0, i = 1, 2, ..., n, \end{aligned}$$

$$(3.7)$$

where $\overline{x}_i(t)$ is the estimation of the *i*th neuron state, $\overline{y}_i(t)$ is the output vector of the *i*th state estimator, $J_i(t)$ is the control input vector, m_i is the estimation gain weight, and m_i is a known constant.

Define the estimation error $r(t) = x(t) - \overline{x}(t)$ and the output error $\tilde{y} = y(t) - \overline{y}(t)$. Then the error state system can be expressed by

$$\dot{r}_{i}(t) = -d_{i}r_{i}(t) + \sum_{j=1}^{n} \overline{a}_{ij}(r_{i}(t))\overline{f}_{j}(r_{j}(t)) + \sum_{j=1}^{n} \overline{b}_{ij}(r_{i}(t))\overline{f}_{j}(r_{j}(\rho_{j}(t))) - m_{i}\overline{f}_{i}(r_{i}(t)) + J_{i}(t),$$

$$\tilde{y}_{i}(t) = \overline{f}_{i}(r_{i}(t)), \quad t \ge 0, i = 1, 2, ..., n,$$
(3.8)

where $\overline{a}_{ij}(r_i(t)) = a_{ij}(x_i(t)) - a_{ij}(\overline{x}_i(t)), \ \overline{b}_{ij}(r_i(t)) = b_{ij}(x_i(t)) - b_{ij}(\overline{x}_i(t)), \ \overline{f}_i(r_i(t)) = f_i(x_i(t)) - f_i(\overline{x}_i(t)), \ \overline{f}_i(r_i(\rho_i(t))) = f_i(x_i(\rho_i(t))) - f_i(\overline{x}_i(\rho_i(t))).$

Define

$$w_i(t) = r_i(e^t), \tag{3.9}$$

then system (3.8) is equivalent to the following neural networks with multiple constant delays:

$$\dot{w}_{i}(t) = e^{t} \left[-d_{i}w_{i}(t) + \sum_{j=1}^{n} \overline{a}_{ij}(w_{i}(t))\overline{f}_{j}(w_{j}(t)) + \sum_{j=1}^{n} \overline{b}_{ij}(w_{i}(t))\overline{f}_{j}(w_{j}(t-\tau_{j})) - m_{i}\overline{f}_{i}(w_{i}(t)) + u_{i}(t) \right],$$

$$\tilde{y}_{i}(t) = \overline{f}_{i}(w_{i}(t)), \quad t \ge 0, i = 1, 2, \dots, n,$$
(3.10)

in which $\tau_{j} = -\log q_{j} \ge 0$, $u_{i}(t) = J_{i}(e^{t})$.

The initial conditions of system (3.10) are assumed to be

$$w_i(t) = \varphi_i(t), \quad t \in [-\tau, 0], \tau = \max_{0 \le j \le n} \{\tau_j\}, i, j = 1, 2, \dots, n_j$$

where $\varphi_i(t) = (\varphi_1(t), \varphi_2(t), \dots, \varphi_n(t)) \in \mathcal{C}([-\tau, 0], \mathbb{R}^n)$, \mathbb{R}^n is *n*-dimensional Euclidean space. $\mathcal{C}([-\rho, 0], \mathbb{R}^n)$ is Banach space of all continuous functions with the Euclidean norm.

Theorem 3.2 Suppose the assumption (A1) holds. If there exists $M = \text{diag}(m_1, m_2, ..., m_n)$ such that

$$\Theta = \begin{pmatrix} \Pi & B_l \\ * & -I \end{pmatrix} < 0, \tag{3.11}$$

where $\Pi = A_l + A_l^T + \Lambda - \Lambda \Gamma - \Gamma^T \Lambda^T - \Lambda M - M^T \Lambda^T$, the estimation error system (3.8) is passive and M is the gain matrix of the state estimator (3.7).

Proof Define a Lyapunov-Krasovskii functional V(x(t)):

$$V(w(t)) = 2\sum_{i=1}^{n} e^{-t} \int_{0}^{w_{i}(t)} \overline{f}_{i}(s) \,\mathrm{d}s + \sum_{i=1}^{n} \int_{t-\tau_{i}}^{t} \overline{f}_{i}^{2}(w_{i}(s)) \,\mathrm{d}s.$$
(3.12)

Calculating the derivative of V(w(t)) along the positive half trajectory of system (3.10), we obtain

$$\begin{split} \dot{V}(w(t)) &- 2\tilde{y}^{T}(t)u(t) - \gamma u^{T}(t)u(t) \\ &= -2\sum_{i=1}^{n} e^{-t} \int_{0}^{w_{i}(t)} \overline{f}_{i}(s) \mathrm{d}s + 2\sum_{i=1}^{n} e^{-t} \overline{f}_{i}(w_{i}(t)) \dot{w}_{i}(t) \\ &+ \sum_{i=1}^{n} \overline{f}_{i}^{2}(w_{i}(t)) - \sum_{i=1}^{n} \overline{f}_{i}^{2}(w_{i}(t-\tau_{i})) - 2\tilde{y}^{T}(t)u(t) - \gamma u^{T}(t)u(t) \end{split}$$

$$\leq 2 \sum_{i=1}^{n} e^{-t} \overline{f}_{i}(w_{i}(t)) \dot{w}_{i}(t) + \sum_{i=1}^{n} \overline{f}_{i}^{2}(w_{i}(t))$$

$$- \sum_{i=1}^{n} \overline{f}_{i}^{2}(w_{i}(t-\tau_{i})) - 2 \overline{y}^{T}(t) u(t) - \gamma u^{T}(t) u(t)$$

$$= 2 \sum_{i=1}^{n} \overline{f}_{i}(w_{i}(t)) \left[-d_{i}w_{i}(t) + \sum_{j=1}^{n} a_{ij}(w_{i}(t)) \overline{f}_{j}(w_{j}(t)) + \sum_{j=1}^{n} b_{ij}(w_{i}(t)) \overline{f}_{j}(w_{j}(t-\tau_{j})) \right]$$

$$- m_{i} \overline{f}_{i}(w_{i}(t)) + u_{i}(t) \right] + \sum_{i=1}^{n} \overline{f}_{i}^{2}(w_{i}(t)) - \sum_{i=1}^{n} \overline{f}_{i}^{2}(w_{i}(t-\tau_{i}))$$

$$- 2 \overline{y}^{T}(t) u(t) - \gamma u^{T}(t) u(t)$$

$$\leq 2 \sum_{i=1}^{n} \left[\left(\frac{1}{2} - \frac{d_{i}}{k_{i}} - m_{i} \right) \overline{f}_{i}^{2}(w_{i}(t)) + \sum_{j=1}^{n} a_{ij}(w_{i}(t)) \overline{f}_{i}(w_{i}(t)) \overline{f}_{j}(w_{j}(t))$$

$$+ \sum_{j=1}^{n} b_{ij}(w_{i}(t)) \overline{f}_{i}(w_{i}(t)) \overline{f}_{j}(w_{j}(t-\tau_{j})) \right] - \sum_{i=1}^{n} \overline{f}_{i}^{2}(w_{i}(t-\tau_{i})) - \gamma u^{T}(t) u(t)$$

$$= 2 \sum_{i=1}^{n} \left[\sum_{j=1}^{n} \left(a_{ij}(w_{i}(t)) + \lambda_{ij} \left(\frac{1}{2} - \frac{d_{i}}{k_{i}} - m_{i} \right) \right) \overline{f}_{i}(w_{i}(t)) \overline{f}_{j}(w_{j}(t))$$

$$+ \sum_{j=1}^{n} b_{ij}(w_{i}(t)) \overline{f}_{i}(w_{i}(t)) \overline{f}_{j}(w_{j}(t-\tau_{j})) \right] - \sum_{i=1}^{n} \overline{f}_{i}^{2}(w_{i}(t-\tau_{i})) - \gamma u^{T}(t) u(t)$$

$$= \eta^{T}(t) \widetilde{\Xi} \eta(t), \qquad (3.13)$$

where

$$\eta(t) = \left[\overline{f}(w(t)), \overline{f}(w(t-\tau)), u(t)\right]^{T}, \qquad \widetilde{\Xi} = \begin{pmatrix} \widetilde{\Pi} & B(w(t)) & 0 \\ * & -I & 0 \\ * & * & -\gamma \end{pmatrix},$$

in which $\widetilde{\Pi} = A(w(t)) + A^T(w(t)) + \Lambda - \Lambda \Gamma - \Gamma^T \Lambda^T - \Lambda M - M^T \Lambda^T$. According to (2.9), we obtain

$$\widetilde{\Xi} = \sum_{l=1}^{2^{2n^2}} \pi_l(t) \begin{pmatrix} \Theta & 0 \\ * & -\gamma \end{pmatrix} \le 0.$$
(3.14)

It follows from (3.13) and (3.14) that

$$\dot{V}(w(t)) - 2y^{T}(t)u(t) - \gamma u^{T}(t)u(t) \le 0.$$
 (3.15)

Integrating (3.15) with respect to t over the time period from 0 to t_p , we have

$$2\int_{0}^{t_{p}} \gamma^{T}(s)u(s) \,\mathrm{d}s \geq V(t_{p}, w(t_{p})) - V(0, w(0)) - \gamma \int_{0}^{t_{p}} u^{T}(t)u(t) \,\mathrm{d}s$$

for w(0) = 0. Since V(0, w(0)) = 0 and $V(t_p, w(t_p)) \ge 0$, we get

$$2\int_0^{t_p} y^T(s)u(s)\,\mathrm{d}s \geq -\gamma \int_0^{t_p} u^T(t)u(t)\,\mathrm{d}s.$$

Therefore, the estimation error system (3.8) is passive in the sense of Definition 5. This completes the proof. $\hfill \Box$

When the control input $J(t) = (J_1(t), J_2(t), \dots, J_n(t))^T = (0, 0, \dots, 0)^T$, by using standard arguments as Theorem 3.2, we can derive the stability conditions for system (3.8).

Corollary 3.2 Suppose the assumption (A1) holds. If there exists $M = \text{diag}(m_1, m_2, ..., m_n)$ such that

$$\Theta = \begin{pmatrix} \Pi & B_l \\ * & -I \end{pmatrix} < 0, \tag{3.16}$$

where $\widetilde{\Pi} = A_l + A_l^T + \Lambda - \Lambda \Gamma - \Gamma^T \Lambda^T - \Lambda M - M^T \Lambda^T$, the estimation error system (3.8) is globally asymptotically stable, while the control input is $J(t) = (J_1(t), J_2(t), \dots, J_n(t))^T = (0, 0, \dots, 0)^T$ and M is the gain matrix of the state estimator (3.7).

4 Numerical simulation

In this section, we give an example to illustrate the feasibility of the theoretical results.

Example 4.1 Consider the two-neuron memristive uncertain neural network model

$$\dot{x}_{1}(t) = -x_{1}(t) + a_{11}(x_{1}(t))f_{1}(x_{1}(t)) + a_{12}(x_{1}(t))f_{2}(x_{2}(t)) + b_{11}(x_{1}(t))f_{1}(x_{1}(\rho_{1}(t))) + b_{12}(x_{1}(t))f_{2}(x_{2}(\rho_{2}(t))) + I_{1}(t), \dot{x}_{2}(t) = -x_{2}(t) + a_{21}(x_{2}(t))f_{1}(x_{1}(t)) + a_{22}(x_{2}(t))f_{2}(x_{2}(t)) + b_{21}(x_{2}(t))f_{1}(x_{1}(\rho_{1}(t))) + b_{22}(x_{2}(t))f_{2}(x_{2}(\rho_{2}(t))) + I_{2}(t),$$

$$(4.1)$$

where $f_1(x) = f_2(x) = \frac{1}{2}(|x+1| - |x-1|)$, $\rho_1(t) = t - (1 - q_1)t$, $\rho_2(t) = t - (1 - q_2)t$, $q_1 = 0.6$, $q_2 = 0.5$, and

$$a_{11}(x_{1}(t)) = \begin{cases} -0.2, & -\frac{df_{1}(x_{1}(t))}{dt} - \frac{dx_{1}(t)}{dt} \le 0, \\ -0.3, & -\frac{df_{1}(x_{1}(t))}{dt} - \frac{dx_{1}(t)}{dt} > 0, \end{cases}$$

$$a_{12}(x_{1}(t)) = \begin{cases} 0.3, & \frac{df_{2}(x_{2}(t))}{dt} - \frac{dx_{1}(t)}{dt} \le 0, \\ 0.7, & \frac{df_{2}(x_{2}(t))}{dt} - \frac{dx_{1}(t)}{dt} > 0, \end{cases}$$

$$b_{11}(x_{1}(t)) = \begin{cases} -0.3, & -\frac{df_{1}(x_{1}(\rho_{1}(t)))}{dt} - \frac{dx_{1}(t)}{dt} \ge 0, \\ -0.4, & -\frac{df_{1}(x_{1}(\rho_{1}(t)))}{dt} - \frac{dx_{1}(t)}{dt} > 0, \end{cases}$$

$$b_{12}(x_{1}(t)) = \begin{cases} 0.4, & \frac{df_{2}(x_{2}(\rho_{2}(t)))}{dt} - \frac{dx_{1}(t)}{dt} \ge 0, \\ 0.8, & \frac{df_{2}(x_{2}(\rho_{2}(t)))}{dt} - \frac{dx_{1}(t)}{dt} \ge 0, \\ 0.5, & \frac{df_{1}(x_{1}(t))}{dt} - \frac{dx_{2}(t)}{dt} \ge 0, \\ 0.5, & \frac{df_{1}(x_{1}(t))}{dt} - \frac{dx_{2}(t)}{dt} \ge 0, \end{cases}$$

$$(4.2)$$

$$a_{22}(x_2(t)) = \begin{cases} -0.2, & -\frac{df_2(x_2(t))}{dt} - \frac{dx_2(t)}{dt} \le 0, \\ -0.8, & -\frac{df_2(x_2(t))}{dt} - \frac{dx_2(t)}{dt} > 0, \end{cases}$$
$$b_{21}(x_2(t)) = \begin{cases} 0.4, & \frac{df_1(x_1(\rho_1(t)))}{dt} - \frac{dx_2(t)}{dt} \le 0, \\ 0.6, & \frac{df_1(x_1(\rho_1(t)))}{dt} - \frac{dx_2(t)}{dt} > 0, \end{cases}$$
$$b_{22}(x_2(t)) = \begin{cases} -0.3, & -\frac{df_2(x_2(\rho_2(t)))}{dt} - \frac{dx_2(t)}{dt} \le 0, \\ -0.9, & -\frac{df_2(x_2(\rho_2(t)))}{dt} - \frac{dx_2(t)}{dt} > 0. \end{cases}$$

In system (4.1), we derive that

$$\Theta = \begin{pmatrix} -1.6 & 0.6 & -0.4 & 0.4 \\ 0.6 & -2.6 & 0.4 & -0.9 \\ * & * & -1 & 0 \\ * & * & * & -1 \end{pmatrix}$$
(4.3)

is negative definite.

By applying Theorem 3.1, the passivity of system (4.1) can be achieved with the input $I_1(t) = 1 + \sin t$, $I_2(t) = 2 + \sin t$. Meanwhile, according to the Corollary 3.1, system (4.1) is globally asymptotically stable with the input $I_1(t) = I_2(t) = 0$. The simulations of the state curves for system (4.1) with and without the input are depicted in Figure 2 and Figure 3, respectively. Obviously, system (4.1) is a state-dependent switched system. By Figure 2, we see that system (4.1) with input $I(t) = (1 + \sin(t), 2 + \sin(t))^T$ can keep internally stable. By adopting the product of input and output as the energy provision, it embodies energy attenuation character. That is, the passive system (4.1) will not produce energy by itself. From Figure 3, it follows that system (4.1) with input $I(t) = (0, 0)^T$ is globally stable. One might see that passivity is at a higher abstraction level of stability by contrast.

We develop a state estimator for neural networks (4.1) as follows:

$$\begin{aligned} \dot{\overline{x}}_{1}(t) &= -\overline{x}_{1}(t) + a_{11}(\overline{x}_{1}(t))f_{1}(\overline{x}_{1}(t)) + a_{12}(\overline{x}_{1}(t))f_{2}(\overline{x}_{2}(t)) \\ &+ b_{11}(\overline{x}_{1}(t))f_{1}(\overline{x}_{1}(\rho_{1}(t))) + b_{12}(\overline{x}_{1}(t))f_{2}(\overline{x}_{2}(\rho_{2}(t))) + I_{1}(t) \\ &+ y_{1}(t) - \overline{y}_{1}(t) - J_{1}(t), \\ \dot{\overline{x}}_{2}(t) &= -\overline{x}_{2}(t) + a_{21}(\overline{x}_{2}(t))f_{1}(\overline{x}_{1}(t)) + a_{22}(\overline{x}_{2}(t))f_{2}(\overline{x}_{2}(t)) \\ &+ b_{21}(\overline{x}_{2}(t))f_{1}(\overline{x}_{1}(\rho_{1}(t))) + b_{22}(\overline{x}_{2}(t))f_{2}(\overline{x}_{2}(\rho_{2}(t))) + I_{2}(t) \\ &+ y_{2}(t) - \overline{y}_{2}(t) - J_{2}(t), \end{aligned}$$

$$(4.4)$$

where $J_1(t) = 1 + \sin(t)$, $J_2(t) = 2 + \sin(t)$ and the other parameters are the same as system (4.1). Then the error state system can be expressed by

$$\dot{r}_{1}(t) = -r_{1}(t) + \overline{a}_{11}(r_{1}(t))\overline{f}_{1}(r_{1}(t)) + \overline{a}_{12}(r_{1}(t))\overline{f}_{2}(r_{2}(t)) + \overline{b}_{11}(r_{1}(t))\overline{f}_{1}(r_{1}(\rho_{1}(t))) + \overline{b}_{12}(r_{1}(t))\overline{f}_{2}(r_{2}(\rho_{2}(t))) - \overline{f}_{1}(r_{1}(t)) + J_{1}(t), \dot{r}_{2}(t) = -r_{2}(t) + \overline{a}_{21}(r_{2}(t))\overline{f}_{1}(r_{1}(t)) + \overline{a}_{22}(r_{2}(t))\overline{f}_{2}(r_{2}(t)) + \overline{b}_{21}(r_{2}(t))\overline{f}_{1}(r_{1}(\rho_{1}(t))) + \overline{b}_{22}(r_{2}(t))\overline{f}_{2}(r_{2}(\rho_{2}(t))) - \overline{f}_{1}(r_{1}(t)) + J_{2}(t).$$

$$(4.5)$$







In system (4.5), we derived

$$\Theta = \begin{pmatrix} -3.6 & 0.6 & -0.4 & 0.4 \\ 0.6 & -4.6 & 0.4 & -0.9 \\ * & * & -1 & 0 \\ * & * & * & -1 \end{pmatrix},$$
(4.6)

which is negative definite.

By applying Theorem 3.2, the passivity of system (4.6) can be achieved with the control input $J_1(t) = 1 + \sin t$, $J_2(t) = 2 + \sin t$ (see Figure 4). Meanwhile, according to the Corollary 3.2, the system (4.6) is globally asymptotically stable with the control input $J_1(t) = J_2(t) = 0$. The responses of the error dynamics (4.5), which converges to zero with the estimation gain matrix K = I can be seen in Figure 5 when the control input J(t) = 0, *i.e.*, the effectiveness of the state estimator for neural networks (4.1) is demonstrated.

It must be pointed out that, in system (4.1), if we let

$$a_{22}(x_2(t)) = \begin{cases} -0.2, & -\frac{df_2(x_2(t))}{dt} - \frac{dx_2(t)}{dt} \le 0, \\ -0.3, & -\frac{df_2(x_2(t))}{dt} - \frac{dx_2(t)}{dt} \ge 0, \end{cases}$$
(4.7)

and the other parameters are the same as above, we obtain

$$\Theta = \begin{pmatrix} -1.6 & 0.6 & -0.4 & 0.4 \\ 0.6 & -0.6 & 0.4 & -0.9 \\ * & * & -1 & 0 \\ * & * & * & -1 \end{pmatrix}.$$







We see that Θ is not negative definite, *i.e.*, the conditions in Theorem 3.1 and Corollary 3.1 are not satisfied. From Figure 6 and Figure 7, we can see that system (4.1) is still passive and stable for the input $I(t) = (1 + \sin(t), 2 + \sin(t))^T$ and $I(t) = (0, 0)^T$, respectively. That is to say, conditions in Theorem 3.1 and Corollary 3.1 are sufficient but not necessary.

5 Concluding remarks

In this paper, we considered a class of memristor-based neural networks with unbounded multiple proportional delays for the designed circuits. By combining differential inclusions with set-valued maps, we obtained the existence of the solutions for the given delayed system. Based on the Lyapunov-Krasovskii functional method, passivity criteria were given and the state estimator was designed for the memristive system through available output measurements based on passivity theory. In system (2.1), we assumed that the decay process of neurons is instantaneous, while, in fact, time is required to isolate the static state. Time delay in the leakage term is introduced to describe the phenomenon which has a great negative impact on the stability of neural networks. On the other hand, distributed delays have caused attention to describe the existence of distribution of conduction velocities along parallel pathways or distribution of propagation delays over a period of time. Accordingly, time delay in the leakage term and distributed delays should be taken into consideration. In the near future, we will analyze memristor-based neural networks with leakage delay and distributed delays, which may reveal more electric properties of the memristor.

The authors declare that they have no competing interests.

Authors' contributions

JL carried out the theoretical calculation and numerical simulation, and drafted the manuscript. RX participated in the theoretical calculation and helped to draft the manuscript. All authors read and approved the final manuscript.

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