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A new method to study ILC problem for time-delay linear systems

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Abstract

In this paper, we apply a new method, a delayed matrix exponential, to study P-type and D-type learning laws for time-delay controlled systems to track the varying reference accurately by using a few iterations in a finite time interval. We present open-loop P- and D-type asymptotic convergence results in the sense of λ -norm by virtue of spectral radius of matrix. Finally, four examples are given to illustrate our theoretical results.

MSC: 34A37; 93C15; 93C40

Keywords: time-delay system; delayed exponential matrix; iterative learning control

1 Introduction

In the past decades, delay differential equations have been widely used in the fields of economics, physics, and engineering control. Existence, stability, and periodic solutions are studied extensively and there are many interesting and important results; see, for example, [1-6].

Since Uchiyama [7] and Arimoto [8] put forward the iterative learning control (ILC for short), a wide variety of iterative learning control problems and related issues are proposed and studied in recent decades. For example, ILC for varying reference trajectories [9, 10], ILC for fractional differential systems [11, 12], ILC for impulsive differential systems [13, 14], research on the robustness of ILC [15, 16], and so on.

It is a common phenomenon that time delays exist in many practical engineering issues. However, the prevalence of the phenomenon to the delay caused a lot of practical engineering problems. So the study of the control problem of time-delay system is paid more attention. Some effective methods for studying the iterative learning control for time-delay systems are provided by Sun [17–19].

After reviewing the previous works dealing with ILC problems for delay systems, we observe the following facts:

- (i) A delay system $\dot{x}(t) = Ax(t) + Bx(t \tau)$, t > 0 is considered mostly as an integral system, where *A*, *B* are suitable matrices.
- (ii) A uniform transition matrix associated with *A*, *B* is not derived directly and the structure of solution *x*(*t*) is not well characterized on every subintervals
 [0, τ],..., [*n*τ, (*n* + 1)τ], *n* ∈ *N*.

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(iii) An extended Gronwall inequality is used to derive convergence results instead of applying direct methods.

It is remarkable that Khusainov and Shuklin in [20] initially introduced a delayed exponential matrix method to study the following linear differential equation with one delay term:

$$\begin{aligned} \dot{x}(t) &= Ax(t) + Bx(t - \tau), \quad t \ge 0, \tau \ge 0, \\ x(t) &= \varphi(t), \quad -\tau \le t \le 0, \end{aligned} \tag{1}$$

where *A* and *B* are matrices and φ is an arbitrary continuously differentiable vector functions. A representation of a solution of system (1) with *AB* = *BA* is given by using a so-called delay exponential matrix, which is defined as follows:

$$e_{\tau}^{Bt} = \begin{cases} \Theta, & t < -\tau, \\ E, & -\tau \le t < 0, \\ E + Bt + \frac{B^2(t-\tau)^2}{2} + \dots + \frac{B^k(t-(k-1)\tau)^k}{k!}, & (k-1)\tau \le t < k\tau, k = 1, 2, \dots, \end{cases}$$
(2)

for $\tau > 0$, where Θ and *E* are the *n*-dimensional zero and identity matrices, respectively.

For more recent contributions on oscillating systems with pure delay, relative controllability of system with pure delay, asymptotic stability of nonlinear multidelay differential equations, one can refer to [21-27] and the references therein.

Inspired by the references mentioned above, in this work, we discuss ILC for time-delay systems. More precisely, we study the following linear controlled systems with pure delay:

$$\begin{cases} \dot{x}_{k}(t) = Ax_{k}(t) + Bx_{k}(t-\tau) + u_{k}(t), & t \in [0, T], \\ x_{k}(t) = \varphi(t), & -\tau \le t \le 0, \tau > 0, \\ y_{k}(t) = Cx_{k}(t) + Du_{k}(t), \end{cases}$$
(3)

where *T* denotes pre-fixed iteration domain length with $T = N\tau$ and $N \in \mathbb{N}$. Let $\varphi \in C_{\tau}^{1} := C^{1}([-\tau, 0], \mathbb{R}^{n})$, *A* and *B* be two $n \times n$ matrices such that AB = BA and *C*, *D* be two $m \times n$ matrices, *k* denotes the *k*th learning iteration, the variables $x_{k}, u_{k} \in \mathbb{R}^{n}$ and $y_{k} \in \mathbb{R}^{m}$ denote state, input, and output, respectively. By [20], Corollary 2.2, we derive that the state $x_{k}(\cdot)$ has the form

$$x_{k}(t) = e^{A(t+\tau)} e_{\tau}^{B_{1}t} \varphi(-\tau) + \int_{-\tau}^{0} e^{A(t-\tau-s)} e_{\tau}^{B_{1}(t-\tau-s)} e^{A\tau} \left[\varphi'(s) - A\varphi(s) \right] ds + \int_{0}^{t} e^{A(t-\tau-s)} e_{\tau}^{B_{1}(t-\tau-s)} e^{A\tau} u_{k}(s) ds,$$
(4)

where e_{τ}^{Bt} is defined in (2) and $B_1 = e^{-A\tau}B$.

By introducing e_{τ}^{Bt} for (2), we state some possible advantages of our approach as follows.

- (i) The structure of solution x(t) is characterized on every subintervals.
- (ii) A direct method is explored to deal with ILC problems by using mathematical analysis tools.

Let y_d be a desired trajectory and set $e_k = y_d - y_k$, which denotes output error and $\delta u_k = u_{k+1} - u_k$.

For the system (3), we consider the open-loop P-type ILC updating law

$$u_{k+1}(t) = u_k(t) + P_o e_k(t).$$
(5)

For the system (3) with $D = \Theta$, we consider the open-loop D-type ILC updating law

$$u_{k+1}(t) = u_k(t) + D_o \dot{e}_k(t), \tag{6}$$

where P_o and $D_o \in \mathbb{R}^{n \times m}$ are learning gain matrices.

The main objective of this paper is to use delayed exponential matrix to generate the control input u_k such that the time-delay system output y_k is tracking the iteratively varying reference trajectories y_d as accurately as possible when $k \to \infty$ uniformly on [0, T] in the sense of the λ -norm by adopting P-type ILC and D-type ILC.

Here we point out that our method is different from the method given in the previous reference, however, we obtain the same convergence results. Our method relies on a direct formula solution, so it is constructive.

The rest of this paper is organized as follows. In Section 2, we give some notations, concepts, and lemmas. In Sections 3 and 4, we give convergence results of P-type ILC and D-type for system (3). Examples are given in Section 5 to demonstrate the applicability of our main results.

2 Preliminaries

Let $J \subset \mathbb{R}$ be a finite interval and $L(\mathbb{R}^n)$ be the space of bounded linear operators in \mathbb{R}^n . Denote by $\mathcal{C}(J, \mathbb{R}^n)$ the Banach space of vector-value continuous functions from $J \to \mathbb{R}^n$ endowed with the ∞ -norm $||x|| = \max_{t \in J} |x(t)|$ for a norm $|\cdot|$ on \mathbb{R}^n . We also consider on $\mathcal{C}(J, \mathbb{R}^n)$ a λ -norm $||x||_{\lambda} = \sup_{t \in J} \{e^{-\lambda t} |x(t)|\}, \lambda > 0$. We introduce a space $\mathcal{C}^1(J, \mathbb{R}^n) =$ $\{x \in \mathcal{C}(J, \mathbb{R}^n) : \dot{x} \in \mathcal{C}(J, \mathbb{R}^n)\}$. For a matrix $A : \mathbb{R}^n \to \mathbb{R}^n$, we consider its matrix norm $||A|| = \max_{|x|=1} |Ax|$ generated by $|\cdot|$.

Lemma 2.1 (see [28], 2.2.8, Chapter 2) Let $A \in L(\mathbb{R}^n)$. For a given $\epsilon > 0$, there is a norm $|\cdot|$ on \mathbb{R}^n such that

 $||A|| \le \rho(A) + \epsilon,$

where $\rho(A)$ denotes the spectral radius of the matrix A.

Next, we give an alternative formula to compute the solution of linear system with pure delay, which is a direct corollary of [20], Corollary 2.2.

Lemma 2.2 Let $f : J \to \mathbb{R}^n$ be a continuous function. The solution $x \in C^1(J, \mathbb{R}^n)$ of

$$\begin{cases} \dot{x}(t) = Ax(t) + Bx(t - \tau) + f(t), & t > 0, \tau \ge 0, \\ x(t) = \varphi(t), & -\tau \le t \le 0, \end{cases}$$
(7)

has the form

$$\begin{aligned} x(t) &= e^{A(t+\tau)} e_{\tau}^{B_{1}t} \varphi(-\tau) + \int_{-\tau}^{0} e^{A(t-\tau-s)} \sum_{l=0}^{j-1} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} \left[\varphi'(s) - A\varphi(s) \right] ds \\ &+ \int_{-\tau}^{t-j\tau} e^{A(t-\tau-s)} B_{1}^{j} \frac{(t-j\tau-s)^{j}}{j!} e^{A\tau} \left[\varphi'(s) - A\varphi(s) \right] ds \\ &+ \sum_{i=0}^{j-1} \int_{0}^{t-i\tau} e^{A(t-\tau-s)} B_{1}^{i} \frac{(t-i\tau-s)^{i}}{i!} e^{A\tau} f(s) ds, \end{aligned}$$
(8)

where A and B are commutative, $B_1 = e^{-A\tau}B$ and $(j-1)\tau \le t < j\tau$, j = 1, 2, ..., N.

Proof According to the formula (21) in [20], we know the solution of system (7) has the form

$$x(t) = e^{A(t+\tau)} e_{\tau}^{B_{1}t} \varphi(-\tau) + \int_{-\tau}^{0} e^{A(t-\tau-s)} e_{\tau}^{B_{1}(t-\tau-s)} e^{A\tau} \left[\varphi'(s) - A\varphi(s) \right] ds + \int_{0}^{t} e^{A(t-\tau-s)} e_{\tau}^{B_{1}(t-\tau-s)} e^{A\tau} f(s) ds.$$
(9)

Without loss of generality, we consider $(j-1)\tau \le t < j\tau$, j = 1, 2, ..., N.

Next, we submit the formula of delayed matrix exponential (2) to (9) to prove the result. We divide our proof into two steps.

Step 1. We prove that

$$\int_{-\tau}^{0} e^{A(t-\tau-s)} e_{\tau}^{B_{1}(t-\tau-s)} e^{A\tau} \left[\varphi'(s) - A\varphi(s) \right] ds$$

$$= \int_{-\tau}^{0} e^{A(t-\tau-s)} \sum_{l=0}^{j-1} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} \left[\varphi'(s) - A\varphi(s) \right] ds$$

$$+ \int_{-\tau}^{t-j\tau} e^{A(t-\tau-s)} B_{1}^{j} \frac{(t-j\tau-s)^{j}}{j!} e^{A\tau} \left[\varphi'(s) - A\varphi(s) \right] ds.$$
(10)

Due to the fact that $-\tau < s < 0$, we obtain $t - \tau < t - \tau - s < t$ and $(j - 2)\tau < t - \tau < t - \tau - s < t < j\tau$. When $-\tau < s < t - j\tau$, we have $(j - 1)\tau < t - \tau - s < t < j\tau$. When $t - j\tau < s < 0$, we have $(j - 2)\tau < t - \tau < s < 0$, we have $(j - 2)\tau < t - \tau < s < 0$.

$$\begin{split} &\int_{-\tau}^{0} e^{A(t-\tau-s)} e_{\tau}^{B_{1}(t-\tau-s)} e^{A\tau} \left[\varphi'(s) - A\varphi(s) \right] ds \\ &= \int_{-\tau}^{t-j\tau} e^{A(t-\tau-s)} \sum_{l=0}^{j} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} \left[\varphi'(s) - A\varphi(s) \right] ds \\ &+ \int_{t-j\tau}^{0} e^{A(t-\tau-s)} \sum_{l=0}^{j-1} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} \left[\varphi'(s) - A\varphi(s) \right] ds \\ &= \int_{-\tau}^{0} e^{A(t-\tau-s)} \sum_{l=0}^{j-1} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} \left[\varphi'(s) - A\varphi(s) \right] ds \\ &+ \int_{-\tau}^{t-j\tau} e^{A(t-\tau-s)} B_{1}^{j} \frac{(t-j\tau-s)^{j}}{j!} e^{A\tau} \left[\varphi'(s) - A\varphi(s) \right] ds. \end{split}$$

Step 2. We check that

$$\int_{0}^{t} e^{A(t-\tau-s)} e^{B_{1}(t-\tau-s)} e^{A\tau} f(s) \, ds = \sum_{i=0}^{j-1} \int_{0}^{t-i\tau} e^{A(t-\tau-s)} B_{1}^{i} \frac{(t-i\tau-s)^{i}}{i!} e^{A\tau} f(s) \, ds. \tag{11}$$

Due to the fact that 0 < s < t, we obtain $-\tau < t - \tau - s < t - \tau$. When $t - (i + 1)\tau < s < t - i\tau$, we have $(i - 1)\tau < t - \tau - s < i\tau$, i = 0, 1, ..., j - 2. When $0 < s < t - (j - 1)\tau$, we have $(j - 2)\tau < t - \tau - s < t - \tau < (j - 1)\tau$. Hence

$$\begin{split} &\int_{0}^{t} e^{A(t-\tau-s)} e^{B_{1}(t-\tau-s)} e^{A\tau} f(s) \, ds \\ &= \sum_{i=0}^{j-2} \int_{t-(i+1)\tau}^{t-i\tau} e^{A(t-\tau-s)} e^{B_{1}(t-\tau-s)} e^{A\tau} f(s) \, ds + \int_{0}^{t-(j-1)\tau} e^{A(t-\tau-s)} e^{B_{1}(t-\tau-s)} e^{A\tau} f(s) \, ds \\ &= \sum_{i=0}^{j-2} \int_{t-(i+1)\tau}^{t-i\tau} e^{A(t-\tau-s)} \sum_{l=0}^{i} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &+ \int_{0}^{t-(j-1)\tau} e^{A(t-\tau-s)} \sum_{l=0}^{j-1} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{i=0}^{j-2} \sum_{l=0}^{i} \int_{t-(i+1)\tau}^{t-i\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &+ \sum_{l=0}^{j-1} \int_{0}^{t-(j-1)\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2} \int_{t-(j-1)\tau}^{t-(j-1)\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2} \int_{0}^{t-i\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2} \int_{0}^{t-(j-1)\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2} \int_{0}^{t-(j-1)\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2} \int_{0}^{t-i\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2} \int_{0}^{t-i\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2} \int_{0}^{t-i\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2} \int_{0}^{t-i\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2} \int_{0}^{t-i\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2} \int_{0}^{t-i\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2} \int_{0}^{t-i\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2} \int_{0}^{t-i\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2} \int_{0}^{t-i\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2} \int_{0}^{t-i\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-l\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2} \int_{0}^{t-i\tau} e^{A(t-\tau-s)} B_{1}^{l} \frac{(t-t\tau-s)^{l}}{l!} e^{A\tau} f(s) \, ds \\ &= \sum_{l=0}^{j-2}$$

Linking (9), (10), and (11), one can get the result (8). The proof is finished.

3 Convergence analysis of P-type

In this section, we give the first convergence result of P-type.

Theorem 3.1 Let $y_d(t)$, $t \in [0, T]$ be a desired trajectory for system (3). If $\rho(E - DP_o) < 1$, then the *P*-type ILC law (5) guarantees $\lim_{k\to\infty} y_k(t) = y_d(t)$ uniformly on [0, T].

Proof Without loss of generality, we consider $(j - 1)\tau \le t < j\tau$, j = 1, 2, ..., N. Linking (3) and (4), we have

$$\begin{aligned} e_{k+1}(t) - e_k(t) &= y_k(t) - y_{k+1}(t) \\ &= -C \int_0^t e^{A(t-\tau-s)} e_\tau^{B_1(t-\tau-s)} e^{A\tau} \delta u_k(s) \, ds - D \delta u_k(t). \end{aligned}$$

According to (5) and Lemma 2.2, we have

$$e_{k+1}(t) = (E - DP_o)e_k(t) - C\sum_{i=0}^{j-1} \int_0^{t-i\tau} e^{A(t-\tau-s)} B_1^i \frac{(t-i\tau-s)^i}{i!} e^{A\tau} \delta u_k(s) \, ds.$$
(12)

Further, by Lemma 2.1 we know that, for a given $\epsilon > 0$, there is a norm $|\cdot|$ on \mathbb{R}^n such that

$$\|E - DP_o\| \le \rho(E - DP_o) + \epsilon.$$

So by (12), we have

$$\begin{aligned} e_{k+1}(t) &| \leq \left[\rho(E - DP_o) + \epsilon \right] \left| e_k(t) \right| \\ &+ \|C\| \sum_{i=0}^{j-1} \int_0^{t-i\tau} \|e^{A(t-\tau-s)}\| \|B_1^i\| \frac{(t-i\tau-s)^i}{i!} \|e^{A\tau}\| |\delta u_k(s)| \, ds. \end{aligned}$$

Hence, we obtain

$$|e_{k+1}(t)| \leq \left[\rho(E - DP_o) + \epsilon\right] |e_k(t)| + \|C\| \sum_{i=0}^{j-1} \int_0^{t-i\tau} e^{\|A\|(t-s)} \|B_1\|^i \frac{(t-i\tau-s)^i}{i!} |\delta u_k(s)| \, ds.$$
(13)

For fixed *i*, i = 0, 1, ..., j - 1, we have

$$\int_{0}^{t-i\tau} e^{\|A\|(t-s)} \|B_{1}\|^{i} \frac{(t-i\tau-s)^{i}}{i!} |\delta u_{k}(s)| ds$$

$$= \frac{\|B_{1}\|^{i}}{i!} \int_{0}^{t-i\tau} e^{\|A\|(t-s)} (t-i\tau-s)^{i} |\delta u_{k}(s)| ds$$

$$\leq \frac{\|B_{1}\|^{i}}{i!} e^{\|A\|t} \int_{0}^{t-i\tau} e^{(\lambda-\|A\|)s} (t-i\tau-s)^{i} ds \|\delta u_{k}\|_{\lambda}.$$
(14)

For any $\lambda > ||A||$, we apply integration by parts via mathematical induction to derive

$$\begin{split} \int_{0}^{t-i\tau} e^{(\lambda - \|A\|)s} (t - i\tau - s)^{i} ds \\ &= \frac{1}{\lambda - \|A\|} \int_{0}^{t-i\tau} (t - i\tau - s)^{i} de^{(\lambda - \|A\|)s} \\ &= -\frac{1}{\lambda - \|A\|} (t - i\tau)^{i} + \frac{i}{(\lambda - \|A\|)^{2}} \int_{0}^{t-i\tau} (t - i\tau - s)^{i-1} de^{(\lambda - \|A\|)s} \end{split}$$

$$= -\frac{1}{\lambda - ||A||} (t - i\tau)^{i} - \frac{i}{(\lambda - ||A||)^{2}} (t - i\tau)^{i-1} + \frac{i(i-1)}{(\lambda - ||A||)^{3}} \int_{0}^{t-i\tau} (t - i\tau - s)^{i-2} de^{(\lambda - ||A||)s} \vdots = -\frac{1}{\lambda - ||A||} (t - i\tau)^{i} - \frac{i}{(\lambda - ||A||)^{2}} (t - i\tau)^{i-1} - \cdots - \frac{i!}{(\lambda - ||A||)^{i}} (t - i\tau) + \frac{i!}{(\lambda - ||A||)^{i+1}} (e^{(\lambda - ||A||)(t-i\tau)} - 1) = -\sum_{p=0}^{i} \frac{i!(t - i\tau)^{i-p}}{(i - p)!(\lambda - ||A||)^{p+1}} + \frac{i!e^{(\lambda - ||A||)(t-i\tau)}}{(\lambda - ||A||)^{i+1}}.$$
(15)

Linking (13), (14), and (15), it is not difficult to get

$$\begin{aligned} |e_{k+1}(t)| &\leq \left[\rho(E - DP_o) + \epsilon\right] |e_k(t)| + ||C|| \sum_{i=0}^{j-1} \frac{||B_1||^i}{i!} e^{||A||t} \\ &\times \left[-\sum_{p=0}^i \frac{i!(t - i\tau)^{i-p}}{(i-p)!(\lambda - ||A||)^{p+1}} + \frac{i!e^{(\lambda - ||A||)(t - i\tau)}}{(\lambda - ||A||)^{i+1}} \right] ||\delta u_k||_{\lambda}. \end{aligned}$$
(16)

By (16) via (5), we have

$$\begin{aligned} \left| e_{k+1}(t) \right| e^{-\lambda t} &\leq \left[\rho(E - DP_o) + \epsilon \right] \left| e_k(t) \right| e^{-\lambda t} + \|C\| \|P_o\| \sum_{i=0}^{j-1} \frac{\|B_1\|^i}{i!} e^{(\|A\| - \lambda)t} \\ &\times \left[-\sum_{p=0}^i \frac{i!(t - i\tau)^{i-p}}{(i-p)!(\lambda - \|A\|)^{p+1}} + \frac{i!e^{(\lambda - \|A\|)(t - i\tau)}}{(\lambda - \|A\|)^{i+1}} \right] \|e_k\|_\lambda \end{aligned}$$

and, taking the λ -norm, we arrive at

$$\begin{split} \|e_{k+1}\|_{\lambda} &\leq \left[\rho(E - DP_o) + \epsilon\right] \|e_k\|_{\lambda} + \|C\| \|P_o\| \sum_{i=0}^{j-1} \frac{\|B_1\|^i}{i!} \\ &\times \left[\sum_{p=0}^i \frac{i! \max\{1, T^p\}}{(i-p)! (\lambda - \|A\|)^{p+1}} + \frac{i!}{(\lambda - \|A\|)^{i+1}}\right] \|e_k\|_{\lambda}. \end{split}$$

Since $\rho(E - DP_o) < 1$, for any $\epsilon \in (0, \frac{1 - \rho(E - DP_o)}{4})$ and $\lambda > ||A||$ sufficiently large, we have

 $\|e_{k+1}\|_{\lambda} \leq \left(\rho(E - DP_o) + 2\epsilon\right)\|e_k\|_{\lambda},$

which implies $\lim_{k\to\infty} \|e_k\|_{\lambda} = 0$, since $\rho(E - DP_o) + 2\epsilon < 1$. In addition, $\|e_k\| \le e^{\lambda T} \|e_k\|_{\lambda}$. Hence $\lim_{k\to\infty} \|e_k\| = 0$. The proof is completed.

Remark 3.2 We use a delayed matrix exponential method to obtain convergence of the P-type ILC algorithm. Next, we applied the norm $\|\cdot\|_{\lambda}$ just for a technical reason to get

uniform convergence only under condition $\rho(E - DP_o) < 1$ in the end of the above proof. Moreover, fixing $\epsilon \in (0, \frac{1-\rho(E-DP_o)}{4})$, the smallest suitable $\lambda > ||A||$ is given by the equation

$$\|C\| \|P_o\| \sum_{i=0}^{N-1} \frac{\|B_1\|^i}{i!} \left[\sum_{p=0}^i \frac{i! \max\{1, T^p\}}{(i-p)! (\lambda - \|A\|)^{p+1}} + \frac{i!}{(\lambda - \|A\|)^{i+1}} \right] = \epsilon,$$
(17)

which is rather awkward to solve. On the other hand, (17) has a unique solution for any $\epsilon > 0$. Moreover, it may have any $\lambda > ||A||$ as a solution by varying $||C|| ||P_o||$.

4 Convergence analysis of D-type

In this section, we discuss the ILC convergence of D-type.

Theorem 4.1 Let $y_d(t)$, $t \in [0, T]$, be a desired trajectory for system (3) with $D = \Theta$. If $\rho(E - CD_o) < 1$ and $e_k(0) = 0$, k = 1, 2, ..., then the *D*-type ILC law (6) guarantees $\lim_{k\to\infty} y_k(t) = y_d(t)$ uniformly on [0, T].

Proof First, we consider $(j-1)\tau \le t < j\tau$, j = 2, 3, ..., N. By (4) and (6), we have

$$\begin{split} \dot{e}_{k+1}(t) - \dot{e}_k(t) &= C \big[\dot{x}_k(t) - \dot{x}_{k+1}(t) \big] \\ &= CA \big[x_k(t) - x_{k+1}(t) \big] + CB \big[x_k(t-\tau) - x_{k+1}(t-\tau) \big] + C \big[u_k(t) - u_{k+1}(t) \big] \\ &= -CA \int_0^t e^{A(t-\tau-s)} e_{\tau}^{B_1(t-\tau-s)} e^{A\tau} \delta u_k(s) \, ds \\ &- CB \int_0^{t-\tau} e^{A(t-2\tau-s)} e_{\tau}^{B_1(t-2\tau-s)} e^{A\tau} \delta u_k(s) \, ds - CD_o \dot{e}_k(t). \end{split}$$

So we have

$$\dot{e}_{k+1}(t) = (E - CD_o)\dot{e}_k(t) - CA \int_0^t e^{A(t-\tau-s)} e_{\tau}^{B_1(t-\tau-s)} e^{A\tau} \delta u_k(s) \, ds - CB \int_0^{t-\tau} e^{A(t-2\tau-s)} e_{\tau}^{B_1(t-2\tau-s)} e^{A\tau} \delta u_k(s) \, ds.$$

Similar to the proof of Theorem 3.1, one can apply Lemma 2.2 to derive that

$$\begin{split} \dot{e}_{k+1}(t) &= (E - CD_o)\dot{e}_k(t) \\ &- CA\sum_{i=1}^j \int_0^{t-(i-1)\tau} e^{A(t-\tau-s)} B_1^{i-1} \frac{(t-(i-1)\tau-s)^{i-1}}{(i-1)!} e^{A\tau} \delta u_k(s) \, ds \\ &- CB\sum_{l=2}^j \int_0^{t-(l-1)\tau} e^{A(t-2\tau-s)} B_1^{l-2} \frac{(t-(l-1)\tau-s)^{l-2}}{(l-2)!} e^{A\tau} \delta u_k(s) \, ds. \end{split}$$

Obviously, we have

$$\begin{aligned} \left| \dot{e}_{k+1}(t) \right| e^{-\lambda t} \\ \leq \left[\rho(E - CD_o) + \epsilon \right] \left| \dot{e}_k(t) \right| e^{-\lambda t} \end{aligned}$$

$$+ \|CA\| \sum_{i=1}^{j} \frac{\|B_{1}\|^{i-1}}{(i-1)!} e^{(\|A\|-\lambda)t} \int_{0}^{t-(i-1)\tau} e^{\|A\|(\lambda-s)} (t-(i-1)\tau-s)^{i-1} ds \|\delta u_{k}\|_{\lambda} + \|CB\| \sum_{l=2}^{j} \frac{\|B_{1}\|^{l-2}}{(l-2)!} e^{(\|A\|-\lambda)t} \int_{0}^{t-(l-1)\tau} e^{\|A\|(\lambda-\tau-s)} (t-(l-1)\tau-s)^{l-2} ds \|\delta u_{k}\|_{\lambda}.$$
(18)

In analogy to the computation in (14)-(16), inequality (18) becomes

$$\left|\dot{e}_{k+1}(t)\right|e^{-\lambda t} \le \left[\rho(E - CD_o) + \epsilon\right] \left|\dot{e}_k(t)\right|e^{-\lambda t} + (W_1 + W_2)\|D_o\|\|\dot{e}_k\|_{\lambda},\tag{19}$$

where

$$W_{1} = \|CA\| \sum_{i=1}^{j} \frac{\|B_{1}\|^{i-1}}{(i-1)!} \left[\sum_{p=1}^{i} \frac{(i-1)!T^{i-p}}{(i-p)!(\lambda - \|A\|)^{p}} + \frac{(i-1)!}{(\lambda - \|A\|)^{i}} \right]$$

and

$$W_{2} = \|CB\| \sum_{l=2}^{j} \frac{\|B_{1}\|^{l-2}}{(l-2)!} \left[\sum_{q=2}^{l} \frac{(l-2)!T^{l-q}}{(l-q)!(\lambda - \|A\|)^{q-1}} + \frac{(l-2)!}{(\lambda - \|A\|)^{l-1}} \right]$$

If j = 1, which means $0 \le t < \tau$, then by (4) and (6), we have

$$\begin{aligned} \dot{e}_{k+1}(t) - \dot{e}_k(t) &= C[\dot{x}_k(t) - \dot{x}_{k+1}(t)] \\ &= CA[x_k(t) - x_{k+1}(t)] + CB[x_k(t-\tau) - x_{k+1}(t-\tau)] + C[u_k(t) - u_{k+1}(t)] \\ &= -CA\int_0^t e^{A(t-\tau-s)} e^{B_1(t-\tau-s)}_\tau e^{A\tau} \delta u_k(s) \, ds - CD_o \dot{e}_k(t). \end{aligned}$$

We can repeat the above arguments to arrive at (19) with $W_2 = 0$ and $W_1 = \frac{2||CA||}{\lambda - ||A||}$. Hence (19) holds on the whole [0, T], which implies

 $\|\dot{e}_{k+1}\|_{\lambda} \leq \left[\rho(E - CD_o) + \epsilon + (W_1 + W_2)\|D_o\|\right] \|\dot{e}_k\|_{\lambda}.$

Note that by $\rho(E - CD_o) < 1$, we obtain $\lim_{k\to\infty} \|\dot{e}_k\|_{\lambda} = 0$. Thus, $\|\dot{e}_k\| \le \dot{e}^{\lambda T} \|\dot{e}_k\|_{\lambda}$. So $\|\dot{e}_k\| \to 0$ as $k \to \infty$. Due to the fact that $e_k(0) = 0$, we get $\|e_k\| \le T \|\dot{e}_k\|$, consequently we find that $\|e_k\| \to 0$ as $k \to \infty$. The proof is completed.

Remark 4.2 Since

$$e_k(0) = y_d(0) - y_k(0) = y_d(0) - Cx_k(0) = y_d(0) - C\varphi(0),$$

we need $y_d(0) = C\varphi(0)$. This gives a compatibility condition for φ . Since $y_d(0)$ is arbitrary, we need *C* to be surjective.

5 Simulation examples

In this section, four numerical examples are presented to demonstrate the validity of the designed method. In order to simulate the tracking errors of trajectories, we adopt for simplicity L^2 -norm in our simulations. This L^2 -norm can be used, since by the fact that $||e_k|| \le$

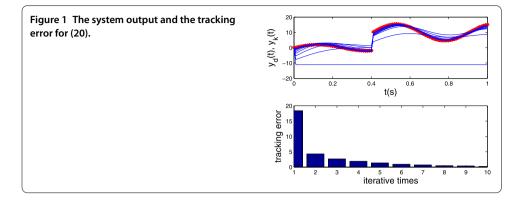


Table 1 The tracking error of each iteration for (20)

k	1	2	3	4	5	6	7	8	9	10
Error	18.41	4.31	2.68	1.91	1.33	0.95	0.68	0.48	0.34	0.24

 $e^{\lambda T} ||e_k||_{\lambda}$ for a suitable $\lambda > ||A||$, we obtain $||e_k|| \to 0$ as $k \to \infty$, *i.e.*, $\lim_{k\to\infty} \sup_{t\in J} |e_k(t)| = 0$, which yields $e_k \to 0$ in $L^2(J, \mathbb{R}^n)$.

Example 5.1 Consider

$$\begin{cases} \dot{x}_{k}(t) = x_{k}(t) + x_{k}(t - 0.5) + u_{k}(t), & x(t) \in \mathbb{R}, t \ge 0, \\ x_{k}(t) = t, & -0.5 \le t \le 0, \\ y_{k}(t) = x_{k}(t) + 0.3u_{k}(t), \end{cases}$$
(20)

and P-type ILC

$$u_{k+1}(t) = u_k(t) + e_k(t).$$

The original reference trajectory is

$$y_d(t) = \begin{cases} 2\sin(4\pi t), & t \in [0, 0.4], \\ 2\sin(4\pi t) + 5\cos(4\pi t) + 10, & t \in [0.4, 1]. \end{cases}$$

Set $t \in [0,1]$, $\tau = 0.5$, $\varphi(t) = t$. For n = m = 1, A = 1, B = 1, C = 1, and D = 0.3. It is not difficult to find that $B_1 = e^{-A\tau}B = e^{-0.5}$ and

$$e_{0.5}^{t} = \begin{cases} 1 + e^{-0.5}t, & t \in [0, 0.5], \\ 1 + e^{-0.5}t + e^{-1}\frac{(t - 0.5)^{2}}{2!}, & t \in [0.5, 1]. \end{cases}$$

Next, we set $P_o = 1$. Obviously, $\rho(1 - DP_o) = 0.7 < 1$. Thus, all conditions of Theorem 3.1 are satisfied, so $y_k(t)$ uniformly converges to $y_d(t)$, for $t \in [0, 1]$.

The upper figure of Figure 1 shows the output y_k of equation (20) of the 10th iterations and the reference trajectory y_d . The lower figure of Figure 1 shows the L^2 -norm of the tracking error (see also Table 1) in each iteration.

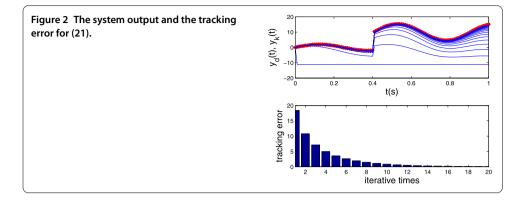


Table 2 The tracking error of each iteration for (21)

	2 10.80				
	12 0.51				

Example 5.2 Consider

$$\begin{cases} \dot{x}_{k}(t) = x_{k}(t) + x_{k}(t - 0.5) + u_{k}(t), & x(t) \in \mathbb{R}, t \ge 0, \\ x_{k}(t) = t, & -0.5 \le t \le 0, \\ y_{k}(t) = x_{k}(t), \end{cases}$$
(21)

and D-type ILC

$$u_{k+1}(t) = u_k(t) + 0.3\dot{e}_k(t)$$

The original reference trajectory is the same as for Example 5.1. Set $t \in [0,1]$, $\tau = 0.5$, $\varphi(t) = t$. For n = m = 1, A = 1, B = 1, C = 1 and D = 0. It is not difficult to find that B_1 and $e_{0.5}^t$ are the same as for Example 5.1. Next, we set $D_o = 0.3$. Obviously, $\rho(1 - CD_o) = 0.7 < 1$. Thus, all conditions of Theorem 4.1 are satisfied.

The upper figure of Figure 2 shows the equation (21) output y_k of the 20th iterations and the reference trajectory y_d . The lower figure of Figure 2 shows the L^2 -norm of the tracking error (see also Table 2) in each iteration.

Example 5.3 Consider

$$\begin{aligned} \dot{x}_{k}(t) &= x_{k}(t) + x_{k}(t - 0.5) + u_{k}(t), \quad x_{k}(t), u_{k}(t) \in \mathbb{R}^{2}, t \ge 0, \\ x_{k}(t) &= (e^{t}, e^{t})^{T}, \quad -0.5 \le t \le 0, \\ y_{k}(t) &= (1, 2)x_{k}(t) + (2, 1)u_{k}(t), \end{aligned}$$
(22)

and P-type ILC

$$u_{k+1}(t) = u_k(t) + \begin{pmatrix} 0.5 \\ -0.4 \end{pmatrix} e_k(t).$$

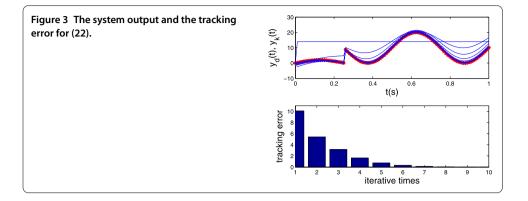


Table 3 The tracking error of each iteration for (22)

k	1	2	3	4	5	6	7	8	9	10
Error	10.09	5.41	3.14	1.62	0.76	0.33	0.14	0.06	0.02	0.01

The original reference trajectory is

$$y_d(t) = \begin{cases} 2\sin(4\pi t), & t \in [0, 0.25], \\ 2\sin(4\pi t) + 8\sin(4\pi t) + 10, & t \in [0.25, 1]. \end{cases}$$

Set $t \in [0,1]$, $\tau = 0.5$, $\varphi(t) = (e^t, e^t)^T$. Next, n = 2, m = 1, A = B = E, E is the identity matrix, C = (1, 2) and D = (2, 1). It is not difficult to find that

$$B_1 = e^{-A\tau} B = \begin{pmatrix} 0.6065 & 0\\ 0 & 0.6065 \end{pmatrix},$$

and

$$e_{0.5}^{t} = \begin{cases} E, & t \in [-0.5, 0], \\ E + B_{1}t, & t \in [0, 0.5], \\ E + B_{1}t + B_{1}^{2} \frac{(t - 0.5)^{2}}{2!}, & t \in [0.5, 1]. \end{cases}$$

Next, we set $P_o = (0.5, -0.4)^T$. Obviously, $\rho(E - DP_o) = 0.8 < 1$. Thus, all conditions of Theorem 3.1 are satisfied. Then $y_k(t)$ uniformly converges to $y_d(t)$, for $t \in [0, 1]$.

The upper figure of Figure 3 shows the equation (22) output y_k of the 10th iterations and the reference trajectory y_d . The lower figure of Figure 3 shows the L^2 -norm of the tracking error (see also Table 3) in each iteration.

Example 5.4 For system

$$\begin{cases} \dot{x}_{k}(t) = x_{k}(t) + x_{k}(t - 0.5) + u_{k}(t), & x_{k}(t), u_{k}(t) \in \mathbb{R}^{2}, t \geq 0, \\ x_{k}(t) = (t, t)^{T}, & -0.5 \leq t \leq 0, \\ y_{k}(t) = (1, 2)x_{k}(t), \end{cases}$$
(23)

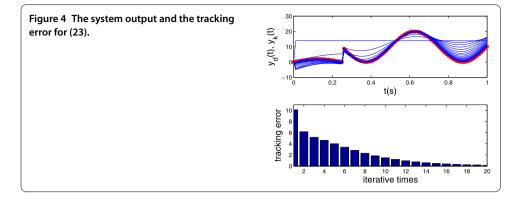


Table 4 The tracking error of each iteration for (23)

				8 2.30	
				18 0.24	

we take the D-type ILC

$$u_{k+1}(t) = u_k(t) + \begin{pmatrix} 1 \\ -0.4 \end{pmatrix} \dot{e}_k(t).$$

The original reference trajectory is the same as for Example 5.3. Obviously, all conditions of Theorem 4.1 are satisfied. Then $y_k(t)$ uniformly converges to $y_d(t)$, for $t \in [0, 1]$.

The upper figure of Figure 4 shows the output y_k of equation (23) of the 20th iterations and the reference trajectory y_d . The lower figure of Figure 4 shows the L^2 -norm of the tracking error (see also Table 4) in each iteration.

Competing interests

The authors declare that they have no competing interests.

Authors' contributions

The authors have made equal contributions. All authors have read and approved the final manuscript.

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Acknowledgements

The authors are grateful to the referees and Associate Professor Dr. Qian Chen for their careful reading of the manuscript and valuable comments. The authors also thank the editor for help. The first and third authors acknowledge National Natural Science Foundation of China (11661016; 11201091), Training Object of High Level and Innovative Talents of Guizhou Province ((2016)4006), United Foundation of Guizhou Province ((2015)7640) and Outstanding Scientific and Technological Innovation Talent Award of Education Department of Guizhou Province ((2014)240). The second author acknowledges the Slovak Grant Agency VEGA No. 1/0078/17 and the Slovak Research and Development Agency under the contract No. APVV-14-0378.

Received: 4 August 2016 Accepted: 6 January 2017 Published online: 31 January 2017

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