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Finite-/fixed-time synchronization of delayed Clifford-valued recurrent neural networks

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Abstract

This paper investigates the problem of finite-/fixed-time synchronization for Clifford-valued recurrent neural networks with time-varying delays. The considered Clifford-valued drive and response system models are firstly decomposed into real-valued drive and response system models in order to overcome the difficulty of the noncommutativity of the multiplication of Clifford numbers. Then, suitable time-delayed feedback controllers are devised to investigate the synchronization problem in finite-/fixed-time of error system. On the basis of new Lyapunov–Krasovskii functional and new computational techniques, finite-/fixed-time synchronization criteria are formulated for the corresponding real-valued drive and response system models. Two numerical examples demonstrate the effectiveness of the theoretical results.

Keywords: Clifford-valued neural network; Synchronization; Finite-time; Fixed-time; Lyapunov–Krasovskii functional

1 Introduction

Recurrent neural network (RNN) models have been used effectively to address optimization, associative memory, signal and image processing, as well as other complex problems. Recently, the dynamic analysis of RNN models has attracted immense interest from various researchers, and useful methods with respect to the stability theory of RNN models have been published [1–4]. In this respect, time delays are frequently observed in neural network (NN) models, owing to the slow speed of signal spread. Time delays are the main source of various dynamics such as chaos, poor functionality, divergence, and instability [5–9]. As such, studies on NN dynamics involving constant or time-varying delays are essential. On the other hand, quaternion-valued and complex-valued NN models are useful in many fields, including night vision color, radar images, polarized signals classification, 3D wind forecasts, and others [10–14]. Recently, many important results concerning different dynamics of the complex-valued and quaternion-valued NN models have been published [6–8, 15–18]. For example, stability analysis [8, 9, 16, 17], finite-time stability [6], stabilizability and instabilizability [7], optimization [19, 20], controllability and observability [21], multistability [18], and so on.

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Clifford algebra provides a solid principle to solve geometry problems. It has been implemented in many diverse areas, e.g., neural systems, computers, robot and control problems [22-26]. Clifford-valued NN models represent a generalization of real-, complex-, and quaternion-valued NN models. To address the challenges associated with highdimensional data and spatial geometric transformation, Clifford-valued NN models are superior to real-, complex-, and quaternion-valued NN models [24–27]. Recently, theoretical and applied studies on Clifford-valued NN models have become a new research subject. However, the dynamic properties of Clifford-valued NN models are usually more complex than those of real-, complex-, and quaternion-valued NN models. Due to the noncommutativity of multiplication with respect to Clifford numbers [28–36], studies on Clifford-valued NN dynamics are still limited. By using the linear matrix inequality approach, the authors in [28] derived the global exponential stability criteria for delayed Clifford RNN models. By using the decomposition process, the issue of global asymptotic stability in Clifford-valued NN models was explored in [29]. In [31], the authors studied the presence of globally asymptotic almost automorphic synchronization to the problem of Clifford-valued RNN models by using suitable feedback controllers. Utilizing the Banach fixed point theorem and Lyapunov functional, the global asymptotic almost periodic synchronization problems for Clifford-valued NN models were examined in [33]. Recently, the effects of neutral delay and discrete delays have been considered in a class of Cliffordvalued NNs [37], and the associated existence, uniqueness, and global stability criteria have been obtained. By considering impulsive effects, the problem of global exponential stability of Clifford-valued NN models with time-varying delays has been investigated in [38].

On the other hand, synchronization is an important dynamical behavior in network systems, which has been applied in many different areas, including image processing, neural computing, traffic systems, and secure communication. As such, various kinds of synchronization have already been suggested in the previous literature [39-46]. Different from the classical synchronization analysis, finite-time synchronization means that the dynamical behaviors of coupled systems achieve the same time spatial state in finite settling time. Therefore, the concept of finite-time synchronization occurs naturally. In recent decades, literature for the finite-time synchronization of NNs has been widely published [47-52]. The use of the feedback controller technique to achieve finite-time synchronization of complex dynamical networks with time delay has been investigated [47]. In [48], the authors showed a detailed study of the finite-time synchronization of stochastic coupled NNs subject to Markovian switching and input saturation. In [49], the authors explored synchronization issues in finite-time complex-valued RNN models with time-varying delays and discontinuous activation functions. With respect to finite-time distributed delays, the synchronization issue of finite-time complex-valued NN models was examined in the work [50].

However, in practical networks, it is difficult to obtain initial values in advance. Therefore, finite-time synchronization of such networks is impossible. Fixed-time synchronization would be a reasonable choice for scientists and engineers in this case as well. The settling time for fixed-time synchronization is calculated using known parameters and does not depend on the initial values, so it is widely accepted when compared to classical finite-time synchronization [53–56]. Furthermore, in realistic engineering problems, it is desirable that the systems can be synchronized for any initial conditions within a fixedtime period. The definition of fixed-time stability was suggested in [53]. Thereafter, the principle of fixed-time stability in various areas, including rigid spacecraft, secure communication, and power systems, was successfully utilized [57–59]. In the papers [54, 55], during fixed-time synchronization of the NNs being addressed, the controllers were designed so as to achieve synchronization criteria within a fixed settling time interval. In [56], a unified control framework was designed for finite-time and fixed-time synchronization of discontinuous complex networks. In [57], fixed-time synchronization of quaternion-valued NNs was obtained by the feedback control principle and the Lyapunov functional method.

However, to the best of authors' knowledge, the problem of finite-/fixed-time synchronization of Clifford-valued RNN with time-varying delays has not yet been considered by any researcher. Indeed, this interesting topic is still an open challenge. Therefore, we attempt to derive sufficient conditions pertaining to finite-/fixed-time synchronization of Clifford-valued RNN models in this paper. The main contributions of this paper are as follows:

- (1) The finite-time and fixed-time synchronization of Clifford-valued RNNs with time-varying delays is investigated for the first time.
- (2) By considering appropriate feedback controllers, Lyapunov functional, and new computational methods, some sufficient conditions that ascertain the finite-/fixed-time synchronization of Clifford-valued RNN models are derived by decomposing the Clifford-valued RNN model into real-valued models.
- (3) When Clifford-valued NN model is reduced to real-, complex-, and quaternion-valued ones, the results obtained in this paper are valid as special cases.
- (4) Two numerical examples with simulations are given to support the effectiveness and merits of the theoretical results.

We organize this paper as follows. The proposed Clifford-valued RNN model is formally defined in Sect. 2. The finite-/fixed-time synchronization criteria are presented in Sect. 3, while two numerical examples are given in Sect. 4. The conclusion of this paper is given in Sect. 5.

2 Mathematical fundamentals and problem formulation

2.1 Notations

The superscripts *T* and * indicate the matrix transposition and matrix involution transposition, respectively. Any matrix $\mathcal{O} > 0$ (< 0) denotes a positive (negative) definite matrix, while \mathbb{A} is defined as the Clifford algebra which has *m* generators over real number \mathbb{R} . In addition, \mathbb{R}^n and \mathbb{A}^n denote the *n*-dimensional real vector space and *n*-dimensional real Clifford vector space, respectively; while $\mathbb{R}^{n \times n}$ and $\mathbb{A}^{n \times n}$ denote the set of all $n \times n$ real matrices and the set of all $n \times n$ real Clifford matrices, respectively. We define the norm of \mathbb{R}^n as $||r|| = \sum_{i=1}^n |r_i|$, and for $A = (a_{ij})_{n \times n} \in \mathbb{R}^{n \times n}$, denote $||A|| = \max_{1 \le i \le n} \{\sum_{j=1}^n |a_{ij}|\}$. While $r = \sum_A r^A e_A \in \mathbb{A}$, denote $|r|_{\mathbb{A}} = \sum_A |r^A|$, and for $A = (a_{ij})_{n \times n} \in \mathbb{A}^{n \times n}$, denote $||A||_{\mathbb{A}} = \max_{1 \le i \le n} \{\sum_{j=1}^n |a_{ij}|\}$. For $\varphi \in C([-\tau, 0], \mathbb{A}^n)$, $||\varphi||_{\tau} \le \sup_{-\tau \le s \le 0} ||\varphi(t + s)||$.

2.2 Clifford algebra

In this subsection, we recall some definitions, notations, and basic results of Clifford algebra. The Clifford real algebra over \mathbb{R}^m is defined as

$$\mathbb{A} = \left\{ \sum_{A \subseteq \{1,2,\dots,m\}} a^A e_A, a^A \in \mathbb{R} \right\},\$$

where $e_A = e_{l_1}e_{l_2}, ..., e_{l_\nu}$ with $A = \{l_1, l_2, ..., l_\nu\}$, $1 \le l_1 < l_2 < \cdots < l_\nu \le m$. Moreover, $e_{\emptyset} = e_0 = 1$ and $e_l = e_{\{l\}}, l = 1, 2, ..., m$, are denoted as the Clifford generators, and they satisfy $e_i e_j + e_j e_i = 0, i \ne j, i, j = 1, 2, ..., m, e_i^2 = -1, i = 1, 2, ..., m$.

Remark 2.1 When one element is the product of multiple Clifford generators, we will write its subscripts together: $e_4e_5e_6e_7 = e_{4567}$.

Let
$$\Lambda = \{\emptyset, 1, 2, \dots, A, \dots, 12, \dots, m\}$$
, and we have

$$\mathbb{A} = \left\{ \sum_{A} a^{A} e_{A}, a^{A} \in \mathbb{R} \right\},$$

where \sum_{A} denotes $\sum_{A \in \Lambda}$ and dim $\mathbb{A} = \sum_{k=0}^{m} {m \choose k} = 2^{m}$. For $r = \sum_{A} r^{A} e_{A} : \mathbb{R} \to \mathbb{A}$, where $r^{A} : \mathbb{R} \to \mathbb{R}, A \in \Lambda$, its derivative is represented by $\frac{dr(t)}{dt} = \sum_{A} \frac{dr^{A}(t)}{dt} e_{A}$. For more knowledge about Clifford algebra, we refer the reader to [28–30].

2.3 Problem definition

Consider the following Clifford-valued RNN model with time-varying delays:

$$\dot{r}_{i}(t) = -d_{i}r_{i}(t) + \sum_{j=1}^{n} a_{ij}h_{j}(r_{j}(t)) + \sum_{j=1}^{n} b_{ij}h_{j}(r_{j}(t-\tau_{j}(t))) + k_{i}, \quad t \ge 0,$$
(1)

$$r_i(t) = \varphi_i(t), \quad t \in [-\tau, 0],$$
 (2)

where $i \in N$, $j \in N$ (N = 1, 2, ..., n), and n corresponds to the number of neurons; $r_i(t) \in \mathbb{A}$ represents the state vector of the *i*th unit; $d_i \in \mathbb{R}^+$ indicates the rate with which the *i*th unit will reset its potential to the resting state in isolation when it is disconnected from the network and external inputs; $a_{ij}, b_{ij} \in \mathbb{A}$ indicate the strengths of connection weights without and with time-varying delays between cells *i* and *j*, respectively; $k_i \in \mathbb{A}$ is an external input on the *i*th unit; $h_j(\cdot) : \mathbb{A}^n \to \mathbb{A}^n$ is the activation function of signal transmission; $\tau_j(t)$ corresponds to the transmission delay which satisfies $0 \le \tau_j(t) \le \tau$, where τ is a positive constant, and $\tau = \max_{1 \le j \le n} \{\tau_j(t)\}$. Furthermore, $\varphi_i \in C([-\tau, 0], \mathbb{A}^n)$ is the initial condition with respect to model (1).

Remark 2.2 It is clear that NN model (1) includes real-valued, complex-valued, and quaternion-valued NN models. These mean that the proposed NN model is more general than the corresponding one in the existing articles. For example, when we consider m = 0 in NN model (1), then the model can be reduced to the real-valued NN model proposed in [5]. If we take m = 1 in NN model (1), then the model can be reduced to the complex-valued NN model proposed in [9]. If we choose m = 2 in NN model (1), then the model can be reduced to the complex-valued NN model proposed in [17].

The corresponding response system is defined by

$$\dot{s}_i(t) = -d_i s_i(t) + \sum_{j=1}^n a_{ij} h_j(s_j(t)) + \sum_{j=1}^n b_{ij} h_j(s_j(t-\tau_j(t))) + u_i(t) + k_i, \quad t \ge 0,$$
(3)

$$s_i(t) = \phi_i(t), \quad t \in [-\tau, 0],$$
 (4)

where $i \in N$, $j \in N$ (N = 1, 2, ..., n), and n corresponds to the number of neurons; $s_i(t) \in \mathbb{A}$ represents the state vector of the *i*th unit; $\phi_i \in C([-\tau, 0], \mathbb{A}^n)$ is the initial condition for model (3). In addition, $u_i(t)$ is state-feedback controller, while other notations associated with (3) and (4) are the same as those in (1) and (2).

(A1) Function $h_j(\cdot)$ fulfills the Lipschitz continuity condition with respect to the *n*-dimensional Clifford vector. For each $j \in N$, there exists a positive constant l_j

such that, for any $x, y \in \mathbb{A}$,

$$\left|h_{j}(x)-h_{j}(y)\right|_{\mathbb{A}}\leq l_{j}|x-y|_{\mathbb{A}},\quad j\in N,$$
(5)

where l_j $(j \in N)$ is known as the Lipschitz constant and $h_j(0) = 0$. In addition, there exists a positive constant l_j such that $|h_j(x)|_{\mathbb{A}} \leq l_j$ for any $x \in \mathbb{A}$.

3 Main results

To address the issue of noncommutativity of multiplication of Clifford numbers, we transform the original Clifford-valued models into multidimensional real-valued models. This can be achieved with the help of $e_A \bar{e}_A = \bar{e}_A e_A = 1$ and $e_B \bar{e}_A e_A = e_B$. Given any $\mathcal{G} \in \mathbb{A}$, unique \mathcal{G}^C that is able to satisfy $\mathcal{G}^C e_C h^A e_A = (-1)^{\sigma[B,\bar{A}]} \mathcal{G}^C h^A e_B = \mathcal{G}^{B,\bar{A}} h^A e_B$ can be identified. By decomposing (1) and (2) into $\dot{r} = \sum_A \dot{r}^A e_A$, we have the following real-valued models:

$$\dot{r}_{i}^{A}(t) = -d_{i}r_{i}^{A}(t) + \sum_{j=1}^{n} \sum_{B} a_{ij}^{A,\bar{B}}h_{j}^{B}(r_{j}(t))$$

$$+ \sum_{j=1}^{n} \sum_{B} b_{ij}^{A,\bar{B}}h_{j}^{B}(r_{j}(t-\tau_{j}(t))) + k_{i}^{A}, \quad t \ge 0,$$

$$r_{i}^{A}(t) = \varphi_{i}^{A}(t), \quad t \in [-\tau, 0],$$
(6)

where

$$\begin{split} r^{A}(t) &= \left(r_{1}^{A}(t), r_{2}^{A}(t), \dots, r_{n}^{A}(t)\right)^{T}, \qquad r_{i}(t) = \sum_{A \in \Lambda} r_{i}^{A}(t)e_{A}, \\ k^{A} &= \left(k_{1}^{A}, k_{2}^{A}, \dots, k_{n}^{A}\right)^{T}, \qquad k_{i} = \sum_{A \in \Lambda} k_{i}^{A}e_{A}, \\ h^{B}(r(t)) &= \left(h_{1}^{B}(r_{1}^{C_{1}}(t), r_{1}^{C_{2}}(t), \dots, r_{1}^{C_{2m}}(t)\right), h_{2}^{B}(r_{2}^{C_{1}}(t), r_{2}^{C_{2}}(t), \dots, r_{2}^{C_{2m}}(t)), \\ & \dots, h_{n}^{B}(r_{n}^{C_{1}}(t), r_{n}^{C_{2}}(t), \dots, r_{n}^{C_{2m}}(t)))^{T}, \\ h^{B}(r(t-\tau(t))) &= \left(h_{1}^{B}(r_{1}^{C_{1}}(t-\tau(t)), r_{1}^{C_{2}}(t-\tau(t)), \dots, r_{1}^{C_{2m}}(t-\tau(t)))\right), \\ & h_{2}^{B}(r_{2}^{C_{1}}(t-\tau(t)), r_{2}^{C_{2}}(t-\tau(t)), \dots, r_{2}^{C_{2m}}(t-\tau(t))), \end{split}$$

$$\dots, h_n^B \left(r_n^{C_1} \left(t - \tau(t) \right), r_n^{C_2} \left(t - \tau(t) \right), \dots, r_n^{C_{2m}} \left(t - \tau(t) \right) \right) \right)^T,$$

$$a_{ij} = \sum_C a_{ij}^C e_C, \qquad a_{ij}^{A,\bar{B}} = (-1)^{\sigma[A,\bar{B}]} a_{ij}^C,$$

$$b_{ij} = \sum_C b_{ij}^C e_C, \qquad b_{ij}^{A,\bar{B}} = (-1)^{\sigma[A,\bar{B}]} b_{ij}^C,$$

$$e_A \bar{e}_B = (-1)^{\sigma[A,\bar{B}]} e_C.$$

Remark 3.1 If $r = (r_1^0, r_1^1, r_1^2, \dots, r_1^{1,2,\dots,m}, r_2^0, r_2^1, r_2^2, \dots, r_2^{1,2,\dots,m}, \dots, r_n^0, r_n^0, r_n^2, \dots, r_n^{1,2,\dots,m})^T = \{r_i^A\}$ is a solution to system (6), then $r = (r_1, r_2, \dots, r_n)^T$ must be a solution to system (1), where $r_i = \sum_A r_i^A e^A$, $i = 1, 2, \dots, n, A \in \Lambda$.

Also, we can use the same method to transform (3) and (4) into the following real-valued models:

$$\dot{s}_{i}^{A}(t) = -d_{i}s_{i}^{A}(t) + \sum_{j=1}^{n}\sum_{B}a_{ij}^{A\bar{B}}h_{j}^{B}(s_{j}(t)) + \sum_{j=1}^{n}\sum_{B}b_{ij}^{A\bar{B}}h_{j}^{B}(s_{j}(t-\tau_{j}(t)))$$

$$+ u_i^A(t) + k_i^A, \quad t \ge 0, \tag{8}$$

$$s_i^A(t) = \phi_i^A(t), \quad t \in [-\tau, 0],$$
(9)

where

$$\begin{split} s^{A}(t) &= \left(s_{1}^{A}(t), s_{2}^{A}(t), \dots, s_{n}^{A}(t)\right)^{T}, \qquad s_{i}(t) = \sum_{A \in \Lambda} s_{i}^{A}(t)e_{A}, \\ u^{A}(t) &= \left(u_{1}^{A}(t), u_{2}^{A}(t), \dots, u_{n}^{A}(t)\right)^{T}, \qquad u_{i}(t) = \sum_{A \in \Lambda} u_{i}^{A}(t)e_{A}, \\ h^{B}(s(t)) &= \left(h_{1}^{B}(s_{1}^{C_{1}}(t), s_{1}^{C_{2}}(t), \dots, s_{1}^{C_{2^{m}}}(t)\right), h_{2}^{B}(s_{2}^{C_{1}}(t), s_{2}^{C_{2}}(t), \dots, s_{2}^{C_{2^{m}}}(t)), \\ &\dots, h_{n}^{B}(s_{n}^{C_{1}}(t), s_{n}^{C_{2}}(t), \dots, s_{n}^{C_{2^{m}}}(t))\right)^{T}, \\ h^{B}(s(t-\tau(t))) &= \left(h_{1}^{B}(s_{1}^{C_{1}}(t-\tau(t)), s_{1}^{C_{2}}(t-\tau(t)), \dots, s_{1}^{C_{2^{m}}}(t-\tau(t)))\right), \\ &h_{2}^{B}(s_{2}^{C_{1}}(t-\tau(t)), s_{2}^{C_{2}}(t-\tau(t)), \dots, s_{n}^{C_{2^{m}}}(t-\tau(t))), \\ &\dots, h_{n}^{B}(s_{n}^{C_{1}}(t-\tau(t)), s_{n}^{C_{2}}(t-\tau(t)), \dots, s_{n}^{C_{2^{m}}}(t-\tau(t))))^{T}. \end{split}$$

Note that the remaining notations of (8) and (9) are the same as those in (6) and (7).

Define the error vector between the real-valued drive models (6), (7) and the real-valued response models (8), (9) as $\mathbf{e}_i^A(t) = s_i^A(t) - r_i^A(t)$ and $\psi_i^A(t) = \phi_i^A(t) - \varphi_i^A(t)$; then, from (6)–(9), the following error models are produced:

$$\dot{\boldsymbol{e}}_{i}^{A}(t) = -d_{i}\boldsymbol{e}_{i}^{A}(t) + \sum_{j=1}^{n} \sum_{B} a_{ij}^{A,\bar{B}} \left(h_{j}^{B} \left(s_{j}(t) \right) - h_{j}^{B} \left(r_{j}(t) \right) \right) \\ + \sum_{j=1}^{n} \sum_{B} b_{ij}^{A,\bar{B}} \left(h_{j}^{B} \left(s_{j} \left(t - \tau_{j}(t) \right) \right) - h_{j}^{B} \left(r_{j} \left(t - \tau_{j}(t) \right) \right) \right) + u_{i}^{A}(t), \quad t \ge 0,$$
(10)

$$\boldsymbol{e}_{i}^{A}(t) = \psi_{i}^{A}(t), \quad t \in [-\tau, 0].$$
(11)

Remark 3.2 As we all know, the multiplication of Clifford numbers does not satisfy the commutative law, which complicates the investigation of Clifford-valued NNs dynamics. Therefore, the known results on Clifford-valued neural networks are limited. On the other hand, the decomposition approach is very effective to overcome the difficulty of the non-commutativity of the multiplication of Clifford numbers. Therefore, it is highly meaningful to use the decomposition method to study Clifford-valued NNs. Recently, most existing results have been derived by decomposing Clifford valued NNs into real-valued NNs [28, 29, 33, 37, 38].

The following lemmas, definitions are used as the main techniques in the paper.

Definition 3.3 ([58]) The origin of a nonlinear dynamical model is said to be globally finite-time stable if, for any solution $r(t, r_0)$, the following are true:

- (1) Lyapunov stability: For any $\epsilon > 0$, there is $\delta = \delta(\epsilon) > 0$ such that $||r(t, r_0)|| < \epsilon$ for any $||r_0|| < \delta$ and $t \ge t_0$.
- (2) Finite-time convergence: There exists a function $T : \mathbb{R}^n \setminus \{0\} \to (0, +\infty)$, denoted as the settling time function, such that $\lim_{t \to T(r_0)} = 0$ and $r(t, r_0) = 0$ for all $t \ge T(r_0)$.

Definition 3.4 ([59]) The origin of the nonlinear dynamical model is said to be fixed-time stable if it is globally finite-time stable and the settling time function $T(r_0)$ is bounded for any $r_0 \in \mathbb{R}^n$, i.e., there exists $T_{\max} > 0$ such that $T(r_0) \le T_{\max}$ for all $r_0 \in \mathbb{R}^n$.

Definition 3.5 ([60]) Consider drive-response models (6) and (7) as well as (8) and (9). If, for a suitable controller $u_i(t) = \sum_A u_i^A(t)e_A$, there exists a function $T = T(\psi_i) > 0$, depending on the initial value ψ_i , such that

$$\lim_{t \to T} \left| s_i^A(t) - r_i^A(t) \right| = \lim_{t \to T} \left| \boldsymbol{e}_i^A(t) \right| = 0$$

and $|\mathbf{e}_i^A(t)| = 0$ for t > T, $i \in N$, $A \in \Lambda$, then drive models (6) and (7) and response models (8) and (9) achieve synchronization in finite-time.

Definition 3.6 ([61]) A function $\mathcal{V}(r) : \mathbb{R}^n \to \mathbb{R}_+$ is *C*-regular if it is

- (1) regular in \mathbb{R}^n ,
- (2) positive definite, i.e., $\mathcal{V}(r) > 0$ for $r \neq 0$ and $\mathcal{V}(0) = 0$,
- (3) radially unbounded, i.e., $\mathcal{V}(r) \to \infty$ as $||r|| \to \infty$.

Definition 3.7 ([57]) Drive-response models (6) and (7) as well as (8) and (9) are said to achieve fixed-time synchronization if, for any initial condition, there exist time T_{max} and a settling time function $T(\boldsymbol{e}_0^A)$ such that

$$\lim_{t \to T(\boldsymbol{e}_0^A)} \boldsymbol{e}^A(t) = 0, \quad \boldsymbol{e}^A(t) = 0, \forall t \ge T(\boldsymbol{e}_0^A),$$

and $T(\boldsymbol{e}_0^A) \leq T_{\max}$, where $\boldsymbol{e}^A(t) = (\boldsymbol{e}_i^A(t))^T$, $\boldsymbol{e}_0^A = (\boldsymbol{e}_i^A(0))^T$, $i \in N, A \in \Lambda$.

Lemma 3.8 ([62]) Suppose that $\mathcal{V} : \mathbb{R}^{2^m n} \to \mathbb{R}_+$ is continuous, differentiable, positive definite, and it satisfies the following differential inequality:

$$\dot{\mathcal{V}}(t) \leq -\mathcal{F}\mathcal{V}^{arphi}(t), \quad t \geq t_0, \mathcal{V}(t_0) \geq 0,$$

where $\mathcal{F} > 0$ and $0 < \upsilon < 1$ are constants. Then, for any given t_0 , $\mathcal{V}(t)$ satisfies the following inequality:

$$\mathcal{V}^{1-\upsilon}(t) \leq \mathcal{V}^{1-\upsilon}(t_0) - \mathcal{F}(1-\upsilon)(t-t_0), \quad t_0 \leq t \leq T^+,$$

and

$$\mathcal{V}(t) = 0, \quad t \geq T^+,$$

and the settling time T^+ is given by

$$T^+ = t_0 + \frac{\mathcal{V}^{1-\upsilon}(t_0)}{\mathcal{F}(1-\upsilon)}.$$

Lemma 3.9 ([59]) If there exists a continuous radically unbounded function $\mathcal{V} : \mathbb{R}^{2^m n} \to \mathbb{R}_+ \cup \{0\}$ such that

- (1) $\mathcal{V}(\mathbf{e}^A) = 0 \Longrightarrow \mathbf{e}^A = 0$,
- (2) If there exist $C_1 > 0$, $C_2 > 0$, 0 < P < 1, and Q > 1 such that

$$\dot{\boldsymbol{\mathcal{V}}}(\boldsymbol{e}^{A}(t)) \leq -\boldsymbol{\mathcal{C}}_{1}\boldsymbol{\mathcal{V}}^{\boldsymbol{\mathcal{P}}}(\boldsymbol{e}^{A}(t)) - \boldsymbol{\mathcal{C}}_{2}\boldsymbol{\mathcal{V}}^{\boldsymbol{\mathcal{Q}}}(\boldsymbol{e}^{A}(t)),$$
(12)

where $\dot{\mathcal{V}}$ denotes the derivative of \mathcal{V} .

Then $\mathcal{V}(\mathbf{e}^{A}(t)) = 0$, $\forall t \geq T_{\max}^{+}$ with the settling time T_{\max}^{+} given by

$$T_{\max}^{+} = \frac{1}{C_{1}(1-\mathcal{P})} + \frac{1}{C_{2}(\mathcal{Q}-1)}.$$
(13)

Lemma 3.10 ([63]) If $\xi_1, \xi_2, \ldots, \xi_n$ are the positive constants and $0 < \alpha_1 < \alpha_2$, then

$$\left(\sum_{i=1}^n \xi_i^{\alpha_2}\right)^{\frac{1}{\alpha_2}} \le \left(\sum_{i=1}^n \xi_i^{\alpha_1}\right)^{\frac{1}{\alpha_1}}$$

3.1 Finite-time synchronization

In this subsection, the finite-time synchronization criterion for the error system models (10) and (11) is derived. The state-feedback controllers for models (10) and (11) are chosen as follows:

$$u_{i}(t) = \sum_{A} u_{i}^{A}(t) e_{A},$$

$$u_{i}^{A}(t) = -\lambda_{1i} \boldsymbol{e}_{i}^{A}(t) - \lambda_{2i} \left| \boldsymbol{e}_{i}^{A}(t) \right|^{\alpha} \operatorname{sgn}(\boldsymbol{e}_{i}^{A}(t)) - \lambda_{3i} \sum_{j=1}^{n} \left| \boldsymbol{e}_{j}^{A}(t - \tau_{j}(t)) \right| \operatorname{sgn}(\boldsymbol{e}_{i}^{A}(t)), \quad (14)$$

where $i \in N$, $j \in N$, $A \in \Lambda$, and $0 < \alpha < 1$ and λ_{1i} , λ_{2i} , λ_{3i} are the parameters that will be determined.

Theorem 3.11 *Based on Assumption (A1) and a proper selection of the parameters to satisfy the following conditions:*

$$d_{i} + \lambda_{1i} - \sum_{i=1}^{n} \sum_{j=1}^{n} \sum_{A} \sum_{B} \left| a_{ij}^{A,\bar{B}} \right| l_{j} \ge 0,$$
(15)

$$\lambda_{3i} - \sum_{i=1}^{n} \sum_{j=1}^{n} \sum_{A} \sum_{B} \left| b_{ij}^{A,\bar{B}} \right| l_{j} \ge 0, \tag{16}$$

$$\lambda_{2i} > 0. \tag{17}$$

Then the error system models (10) and (11) can achieve the finite-time synchronization with controller (14). Moreover, the settling time of synchronization T^+ satisfies

$$T^{+} = t_{0} + \frac{\mathcal{V}^{1-\alpha}(t_{0})}{\min_{i \in \mathcal{N}}(\lambda_{2i})(1-\alpha)}.$$
(18)

Proof Consider the following Lyapunov function which is positive definite and radially unbounded:

$$\mathcal{V}(t) = \sum_{i=1}^{n} \sum_{A} \left| \boldsymbol{e}_{i}^{A}(t) \right|.$$
(19)

The Dini derivative $\mathcal{D}^+\mathcal{V}(t)$ is computed with model (10). We derive

$$\begin{aligned} \mathcal{D}^{+}\mathcal{V}(t) &= \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t))\dot{\mathbf{e}}_{i}^{A}(t) \\ &= \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \left(-d_{i}\mathbf{e}_{i}^{A}(t) + \sum_{j=1}^{n} \sum_{B} a_{ij}^{A,\bar{B}}(h_{j}^{B}(s_{j}(t)) - h_{j}^{B}(r_{j}(t))) \right) \\ &+ \sum_{j=1}^{n} \sum_{B} b_{ij}^{A,\bar{B}}(h_{j}^{B}(s_{j}(t-\tau_{j}(t))) - h_{j}^{B}(r_{j}(t-\tau_{j}(t)))) + u_{i}^{A} \right) \\ &= \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \left(-d_{i}\mathbf{e}_{i}^{A}(t) + \sum_{j=1}^{n} \sum_{B} a_{ij}^{A,\bar{B}}(h_{j}^{B}(s_{j}(t)) - h_{j}^{B}(r_{j}(t))) \right) \\ &+ \sum_{j=1}^{n} \sum_{B} b_{ij}^{A,\bar{B}}(h_{j}^{B}(s_{j}(t-\tau_{j}(t))) - h_{j}^{B}(r_{j}(t-\tau_{j}(t)))) - \lambda_{1i}\mathbf{e}_{i}^{A}(t) \\ &- \lambda_{2i} |\mathbf{e}_{i}^{A}(t)|^{\alpha} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) - \lambda_{3i} \sum_{j=1}^{n} |\mathbf{e}_{j}^{A}(t-\tau_{j}(t))| \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \right) \\ &= -\sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) d_{i}\mathbf{e}_{i}^{A}(t) \\ &+ \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \sum_{j=1}^{n} \sum_{B} a_{ij}^{A,\bar{B}}(h_{j}^{B}(s_{j}(t-\tau_{j}(t))) - h_{j}^{B}(r_{j}(t-\tau_{j}(t))) \right) \end{aligned}$$

$$-\sum_{i=1}^{n}\sum_{A}\operatorname{sgn}(\boldsymbol{e}_{i}^{A}(t))\lambda_{1i}\boldsymbol{e}_{i}^{A}(t) - \sum_{i=1}^{n}\sum_{A}\operatorname{sgn}(\boldsymbol{e}_{i}^{A}(t))\lambda_{2i}|\boldsymbol{e}_{i}^{A}(t)|^{\alpha}\operatorname{sgn}(\boldsymbol{e}_{i}^{A}(t))$$

$$-\sum_{i=1}^{n}\sum_{j=1}^{n}\sum_{A}\operatorname{sgn}(\boldsymbol{e}_{i}^{A}(t))\lambda_{3i}|\boldsymbol{e}_{j}^{A}(t-\tau_{j}(t))|\operatorname{sgn}(\boldsymbol{e}_{i}^{A}(t))$$

$$=\sum_{i=1}^{n}\sum_{A}-d_{i}|\boldsymbol{e}_{i}^{A}(t)| + \sum_{i=1}^{n}\sum_{A}\operatorname{sgn}(\boldsymbol{e}_{i}^{A}(t))\sum_{j=1}^{n}\sum_{B}d_{ij}^{A\bar{B}}(h_{j}^{B}(s_{j}(t)) - h_{j}^{B}(r_{j}(t)))$$

$$+\sum_{i=1}^{n}\sum_{A}\operatorname{sgn}(\boldsymbol{e}_{i}^{A}(t))\sum_{j=1}^{n}\sum_{B}b_{ij}^{A\bar{B}}(h_{j}^{B}(s_{j}(t-\tau_{j}(t))) - h_{j}^{B}(r_{j}(t-\tau_{j}(t))))$$

$$-\sum_{i=1}^{n}\sum_{A}\lambda_{1i}|\boldsymbol{e}_{i}^{A}(t)| - \sum_{i=1}^{n}\sum_{A}\lambda_{2i}|\boldsymbol{e}_{i}^{A}(t)|^{\alpha} - \sum_{i=1}^{n}\sum_{j=1}^{n}\sum_{A}\lambda_{3i}|\boldsymbol{e}_{j}^{A}(t-\tau_{j}(t))|$$

$$=\sum_{i=1}^{n}\sum_{A}-d_{i}|\boldsymbol{e}_{i}^{A}(t)| + \sum_{i=1}^{n}\sum_{j=1}^{n}\sum_{A}\sum_{B}|d_{ij}^{A\bar{B}}||h_{j}^{B}(s_{j}(t-\tau_{j}(t))) - h_{j}^{B}(r_{j}(t-\tau_{j}(t)))|$$

$$+\sum_{i=1}^{n}\sum_{j=1}^{n}\sum_{A}\sum_{B}|b_{ij}^{A\bar{B}}||h_{j}^{B}(s_{j}(t-\tau_{j}(t))) - h_{j}^{B}(r_{j}(t-\tau_{j}(t)))|$$

$$=\sum_{i=1}^{n}\sum_{A}-d_{i}|\boldsymbol{e}_{i}^{A}(t)| - \sum_{i=1}^{n}\sum_{A}\sum_{B}|\boldsymbol{e}_{i}^{A}(t)|^{\alpha} - \sum_{i=1}^{n}\sum_{j=1}^{n}\sum_{A}\lambda_{3i}|\boldsymbol{e}_{j}^{A}(t-\tau_{j}(t))|$$

$$+\sum_{i=1}^{n}\sum_{j=1}^{n}\sum_{A}\sum_{B}|b_{ij}^{A\bar{B}}||h_{j}^{B}(s_{j}(t-\tau_{j}(t))| - h_{j}^{B}(r_{j}(t-\tau_{j}(t)))|$$

$$\leq\sum_{i=1}^{n}\sum_{A}-d_{i}|\boldsymbol{e}_{i}^{A}(t)| + \sum_{i=1}^{n}\sum_{j=1}^{n}\sum_{A}\sum_{B}|\boldsymbol{e}_{i}^{A\bar{B}}||l_{j}|\boldsymbol{e}_{j}^{A}(t)|^{\alpha} - \sum_{i=1}^{n}\sum_{A}\lambda_{3i}|\boldsymbol{e}_{j}^{A}(t-\tau_{j}(t))|$$

$$\leq\sum_{i=1}^{n}\sum_{A}\lambda_{1i}|\boldsymbol{e}_{i}^{A}(t)| + \sum_{i=1}^{n}\sum_{j=1}^{n}\sum_{A}\lambda_{2i}|\boldsymbol{e}_{i}^{A}(t-\tau_{j}(t))| - \sum_{i=1}^{n}\sum_{A}\lambda_{3i}|\boldsymbol{e}_{j}^{A}(t-\tau_{j}(t))|$$

$$=\sum_{i=1}^{n}\sum_{A}\lambda_{2i}|\boldsymbol{e}_{i}^{A}(t)|^{\alpha} - \sum_{i=1}^{n}\sum_{A}\lambda_{3i}|\boldsymbol{e}_{j}^{A}(t-\tau_{j}(t))|.$$

$$(20)$$

By combining similar terms, we have

$$\mathcal{D}^{+}\mathcal{V}(t) \leq \sum_{i=1}^{n} \sum_{A} \left(-d_{i} - \lambda_{1i} + \sum_{j=1}^{n} \sum_{B} \left| a_{ij}^{A,\bar{B}} \right| l_{j} \right) \left| \mathbf{e}_{i}^{A}(t) \right|$$

$$+ \sum_{i=1}^{n} \sum_{A} \left(-\lambda_{3i} + \sum_{j=1}^{n} \sum_{B} \left| b_{ij}^{A,\bar{B}} \right| l_{j} \right) \left| \mathbf{e}_{j}^{A}(t - \tau_{j}(t)) \right| - \sum_{i=1}^{n} \sum_{A} \lambda_{2i} \left| \mathbf{e}_{i}^{A}(t) \right|^{\alpha}$$

$$\leq - \sum_{i=1}^{n} \sum_{A} \lambda_{2i} \left| \mathbf{e}_{i}^{A}(t) \right|^{\alpha}$$

$$\leq - \min_{i \in N} (\lambda_{2i}) \sum_{i=1}^{n} \sum_{A} \left| \mathbf{e}_{i}^{A}(t) \right|^{\alpha}. \tag{21}$$

From Lemma (3.10), we obtain

$$\sum_{i=1}^{n} \sum_{A} \left| e_i^A(t) \right|^{\alpha} \le \left(\sum_{i=1}^{n} \sum_{A} \left| \boldsymbol{e}_i^A(t) \right| \right)^{\alpha}.$$
(22)

Replacing (22) in (21), we have

$$\mathcal{D}^{+}\mathcal{V}(t) \leq -\min(\lambda_{2i}) \left(\sum_{i=1}^{n} \sum_{A} \left| \boldsymbol{e}_{i}^{A}(t) \right| \right)^{\alpha},$$

$$= -\min_{i \in N} (\lambda_{2i}) \mathcal{V}^{\alpha}(t).$$
(23)

According to Lemma (3.8), response model (8) with controller (14) can be synchronized in finite-time with drive model (6). Furthermore, the settling time of synchronization T^+ is given by (18). Thus, Theorem 3.11 is proved.

3.2 Fixed-time synchronization

The fixed-time synchronization criterion for drive-response models (10) and (11) is derived. The time-delay feedback controller for model (10) is chosen as follows:

$$u_{i}(t) = \sum_{A} u_{i}^{A}(t)e_{A},$$

$$u_{i}^{A}(t) = -\lambda_{1i}\boldsymbol{e}_{i}^{A}(t) - \lambda_{2i} \left|\boldsymbol{e}_{i}^{A}(t)\right|^{\alpha} \operatorname{sgn}\left(\boldsymbol{e}_{i}^{A}(t)\right) - \lambda_{3i} \left|\boldsymbol{e}_{i}^{A}(t)\right|^{\beta} \operatorname{sgn}\left(\boldsymbol{e}_{i}^{A}(t)\right)$$

$$-\lambda_{4i} \left|\sum_{j=1}^{n} \boldsymbol{e}_{j}^{A}\left(t - \tau_{j}(t)\right)\right| \operatorname{sgn}\left(\boldsymbol{e}_{i}^{A}(t)\right),$$
(24)

where $i \in N$, $j \in N$, $A \in \Lambda$, and $0 < \alpha < 1$, $\beta > 1$ and λ_{1i} , λ_{2i} , $\lambda_{3i} \lambda_{4i}$ are the parameters that will be determined.

Theorem 3.12 *Based on Assumption* (A1) *and a proper selection of the parameters to satisfy the following conditions:*

$$d_{i} + \lambda_{1i} - \sum_{i=1}^{n} \sum_{j=1}^{n} \sum_{A} \sum_{B} \left| a_{ij}^{A,\bar{B}} \right| l_{j} \ge 0,$$
(25)

$$\lambda_{4i} - \sum_{i=1}^{n} \sum_{j=1}^{n} \sum_{A} \sum_{B} \left| b_{ij}^{A,\bar{B}} \right| l_{j} \ge 0,$$
(26)

$$\lambda_{2i} > 0, \qquad \lambda_{3i} > 0. \tag{27}$$

The error system models (10) *and* (11) *can achieve the fixed-time synchronization with controller* (24). *Moreover, the settling time of synchronization* T^+ *satisfies*

$$T_{\max}^{+} = \frac{1}{\Pi_2(1-\alpha)} + \frac{1}{\Pi_1(\beta-1)}.$$
(28)

Proof Consider the same Lyapunov function defined in Theorem 3.11:

$$\mathcal{V}(t) = \sum_{i=1}^{n} \sum_{A} \left| \boldsymbol{e}_{i}^{A}(t) \right|.$$
(29)

The Dini derivative $\mathcal{D}^+\mathcal{V}(t)$ is computed with model (10). We derive

$$\begin{split} \mathcal{D}^{*}\mathcal{V}(t) &= \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \dot{\mathbf{e}}_{i}^{A}(t) \\ &= \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \left(-d_{i}\mathbf{e}_{i}^{A}(t) + \sum_{j=1}^{n} \sum_{B} a_{ij}^{A,B}(h_{j}^{B}(s_{j}(t)) - h_{j}^{B}(r_{j}(t))) \right) \\ &+ \sum_{j=1}^{n} \sum_{B} b_{ij}^{A,B}(h_{j}^{B}(s_{j}(t-\tau_{j}(t))) - h_{j}^{B}(r_{j}(t-\tau_{j}(t)))) + u_{i}^{A} \right) \\ &= \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \left(-d_{i}\mathbf{e}_{i}^{A}(t) + \sum_{j=1}^{n} \sum_{B} a_{ij}^{A,B}(h_{j}^{B}(s_{j}(t)) - h_{j}^{B}(r_{j}(t))) \right) \\ &+ \sum_{j=1}^{n} \sum_{B} b_{ij}^{A,B}(h_{j}^{B}(s_{j}(t-\tau_{j}(t)))) - h_{j}^{B}(r_{j}(t-\tau_{j}(t)))) - \lambda_{1i}\mathbf{e}_{i}^{A}(t) \\ &- \lambda_{2i}|\mathbf{e}_{i}^{A}(t)|^{\alpha} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) - \lambda_{3i}|\mathbf{e}_{i}^{A}(t)|^{\beta} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \\ &- \lambda_{4i}\sum_{j=1}^{n} |\mathbf{e}_{j}^{A}(t-\tau_{j}(t))| \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \right) \\ &= -\sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) d_{i}\mathbf{e}_{i}^{A}(t) \\ &+ \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) d_{i}\mathbf{e}_{i}^{A}(t) \\ &+ \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \lambda_{1i}\mathbf{e}_{i}^{A}(t) - \sum_{j=1}^{n} \sum_{B} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) - h_{j}^{B}(r_{j}(t-\tau_{j}(t))) \\ &- \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \lambda_{1i}\mathbf{e}_{i}^{A}(t) - \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \lambda_{2i}|\mathbf{e}_{i}^{A}(t)|^{\alpha} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \\ &- \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \lambda_{3i}|\mathbf{e}_{i}^{A}(t)|^{\beta} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \\ &- \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \lambda_{3i}|\mathbf{e}_{i}^{A}(t)|^{\beta} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \\ &- \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \lambda_{3i}|\mathbf{e}_{i}^{A}(t)|^{\beta} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \\ &+ \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \sum_{j=1}^{n} \sum_{B} a_{ij}^{A,B}(h_{j}^{B}(s_{j}(t-\tau_{j}(t))) - h_{j}^{B}(r_{j}(t-\tau_{j}(t)))) \\ &+ \sum_{i=1}^{n} \sum_{A} \operatorname{sgn}(\mathbf{e}_{i}^{A}(t)) \sum_{j=1}^{n} \sum_{B} b_{ij}^{A,B}(h_{j}^{B}(s_{j}(t) - \tau_{j}(t))) - h_{j}^{B}(r_{j}(t-\tau_{j}(t))))) \\ &- \sum_{i=1}^{n} \sum_{A} \lambda_{1i}|\mathbf{e}_{i}^{A}(t)| - \sum_{i=1}^{n} \sum_{B} \lambda_{2i}|\mathbf{e}_{i}^{A}(t)|^{\alpha} - \sum_{i=1}^{n} \sum_{A} \lambda_{3i}|\mathbf{e}_{i}^{A}(t)|^{\beta} \end{split}$$

$$-\sum_{i=1}^{n}\sum_{j=1}^{n}\sum_{A}^{n}\lambda_{4i}|\mathbf{e}_{i}^{A}(t-\tau_{j}(t))|$$

$$=\sum_{i=1}^{n}\sum_{A}^{n}-d_{i}|\mathbf{e}_{i}^{A}(t)|+\sum_{i=1}^{n}\sum_{j=1}^{n}\sum_{A}^{n}\sum_{B}|a_{ij}^{A,\bar{B}}||h_{j}^{B}(s_{j}(t))-h_{j}^{B}(r_{j}(t))|$$

$$+\sum_{i=1}^{n}\sum_{j=1}^{n}\sum_{A}\sum_{B}|b_{ij}^{A,\bar{B}}||h_{j}^{B}(s_{j}(t-\tau_{j}(t)))-h_{j}^{B}(r_{j}(t-\tau_{j}(t)))|$$

$$-\sum_{i=1}^{n}\sum_{A}\lambda_{1i}|\mathbf{e}_{i}^{A}(t)|-\sum_{i=1}^{n}\sum_{A}\lambda_{2i}|\mathbf{e}_{i}^{A}(t)|^{\alpha}-\sum_{i=1}^{n}\sum_{A}\lambda_{3i}|\mathbf{e}_{i}^{A}(t)|^{\beta}$$

$$-\sum_{i=1}^{n}\sum_{j=1}^{n}\sum_{A}\lambda_{4i}|\mathbf{e}_{i}^{A}(t)|+\sum_{i=1}^{n}\sum_{j=1}^{n}\sum_{A}\sum_{B}|a_{ij}^{A,\bar{B}}|l_{j}|\mathbf{e}_{j}^{A}(t)|$$

$$+\sum_{i=1}^{n}\sum_{j=1}^{n}\sum_{A}\sum_{B}|b_{ij}^{A,\bar{B}}|l_{j}|\mathbf{e}_{j}^{A}(t-\tau_{j}(t))|-\sum_{i=1}^{n}\sum_{A}\lambda_{1i}|\mathbf{e}_{i}^{A}(t)|$$

$$-\sum_{i=1}^{n}\sum_{A}\lambda_{2i}|\mathbf{e}_{i}^{A}(t)|^{\alpha}-\sum_{i=1}^{n}\sum_{A}\lambda_{3i}|\mathbf{e}_{i}^{A}(t)|^{\beta}$$

$$-\sum_{i=1}^{n}\sum_{A}\lambda_{2i}|\mathbf{e}_{i}^{A}(t)|^{\alpha}-\sum_{i=1}^{n}\sum_{A}\lambda_{3i}|\mathbf{e}_{i}^{A}(t)|^{\beta}$$

$$(30)$$

By combining similar terms, we have

$$\mathcal{D}^{+}\mathcal{V}(t) \leq \sum_{i=1}^{n} \sum_{A} \left(-d_{i} - \lambda_{1i} + \sum_{j=1}^{n} \sum_{B} \left| a_{ij}^{A,\bar{B}} \right| l_{j} \right) \left| \mathbf{e}_{i}^{A}(t) \right| \\ + \sum_{i=1}^{n} \sum_{A} \left(-\lambda_{4i} + \sum_{j=1}^{n} \sum_{B} \left| b_{ij}^{A,\bar{B}} \right| l_{j} \right) \left| \mathbf{e}_{j}^{A}(t - \tau_{j}(t)) \right| \\ - \sum_{i=1}^{n} \sum_{A} \lambda_{2i} \left| \mathbf{e}_{i}^{A}(t) \right|^{\alpha} - \sum_{i=1}^{n} \sum_{A} \lambda_{3i} \left| \mathbf{e}_{i}^{A}(t) \right|^{\beta} \\ \leq -\sum_{i=1}^{n} \sum_{A} \lambda_{2i} \left| \mathbf{e}_{i}^{A}(t) \right|^{\alpha} - \sum_{i=1}^{n} \sum_{A} \lambda_{3i} \left| \mathbf{e}_{i}^{A}(t) \right|^{\beta}.$$
 (31)

From Lemma (3.10), we obtain

$$\sum_{i=1}^{n} \sum_{A} \lambda_{2i} \left| \boldsymbol{e}_{i}^{A}(t) \right|^{\alpha} \leq \left(\sum_{i=1}^{n} \sum_{A} \lambda_{2i} \left| \boldsymbol{e}_{i}^{A}(t) \right| \right)^{\alpha}, \tag{32}$$

$$\sum_{i=1}^{n} \sum_{A} \lambda_{3i} \left| \boldsymbol{e}_{i}^{A}(t) \right|^{\beta} \leq \left(\sum_{i=1}^{n} \sum_{A} \lambda_{3i} \left| \boldsymbol{e}_{i}^{A}(t) \right| \right)^{\beta}.$$
(33)

Replacing (32) and (33) in (31), we have

$$\mathcal{D}^{+}\mathcal{V}(t) \leq -\left(\sum_{i=1}^{n}\sum_{A}\lambda_{2i}\left|\boldsymbol{e}_{i}^{A}(t)\right|\right)^{\alpha} - \left(\sum_{i=1}^{n}\sum_{A}\lambda_{3i}\left|\boldsymbol{e}_{i}^{A}(t)\right|\right)^{\beta}$$
$$\leq -\left[\min_{i\in\mathcal{N}}(\lambda_{2i})\right]^{\alpha}\mathcal{V}^{\alpha}(t) - \left[\min_{i\in\mathcal{N}}(\lambda_{3i})\right]^{\beta}\mathcal{V}^{\beta}(t).$$
(34)

According to Lemma (3.9), response model (8) can be synchronized in fixed time with drive model (6) under controller (24). The settling time of synchronization T^+_{\max} max is given by (28) with $\Pi_1 = [\min_{i \in N} (\lambda_{3i})]^{\beta}$, $\Pi_2 = [\min_{i \in N} (\lambda_{2i})]^{\alpha}$. Thus, Theorem 3.12 is proved.

Remark 3.13 In Theorem 3.11 and Theorem 3.12, by decomposing the original *n*-dimensional Clifford-valued system into a multidimensional real-valued system, several sufficient conditions are derived to show that the considered system model is finite-time and fixed-time synchronized, but the result we get is really about Clifford-valued systems themselves.

Remark 3.14 As everyone knows, Clifford-valued neural networks aim to investigate new capabilities and better accuracy in order to solve problems that cannot be solved with real-valued, complex-valued, quaternion-valued counterparts. For example, the results of finite-time and fixed-time synchronization of complex-valued NNs [49], fixed-time synchronization of quaternion-valued NNs [57] can then be summarized as a special case of the results of this paper.

Remark 3.15 In [30], fuzzy operations are incorporated into the Clifford-valued cellular NN model to investigate its S^p -almost periodic solutions. The effects of discrete delays in Clifford-valued recurrent NNs are considered in [31], and the associated globally asymptotic almost automorphic synchronization criteria are obtained. The leakage delay is introduced into Clifford-valued high-order Hopfield NN models in [32] to explore its existence and global exponential stability of almost automorphic solutions. However, any work on the topic of finite-/fixed-time synchronization of Clifford-valued RNN with time-varying delays has not yet been reported. Therefore, trying to fill such gaps, we for the first time derived new sufficient conditions to ensure the finite/fixed-time synchronization of Clifford-valued RNN models with time delays. Therefore, the main results of this paper are new and different compared with those in the existing literature.

Remark 3.16 The computational complexity depends primarily on the maximum number of LMI decision variables. As everyone knows, the number of decision variables increases when using the augmented Lyapunov–Krasovskii functionals and the free-weightingmatrix method; while when the delay subintervals number becomes higher, it might prompt the complexity and the computational burden of the main results. In order to handle this issue easily, we utilized standard Lyapunov functional and estimated its time derivative without any integral inequalities and delay-decomposition approach. Hence, the proposed results in this paper may provide a smaller computational burden. Remark 3.17 Compared with many controllers derived by early works [45, 46], the feedback controller methods are economic and easy to implement, they possess high value in real industry processes and applications.

4 Numerical examples

In this section, we present two numerical examples to demonstrate the feasibility and effectiveness of the results established in Sect. 3.

Example 1 For m = 2 and n = 2, the following two-neuron drive model (1) is considered:

$$\begin{split} \dot{r}_1(t) &= -d_1 r_1(t) + a_{11} h_1 \big(r_1(t) \big) + a_{12} h_2 \big(r_2(t) \big) + b_{11} h_1 \big(r_1 \big(t - \tau_1(t) \big) \big) \\ &+ b_{12} h_2 \big(r_2 \big(t - \tau_2(t) \big) \big) + k_1, \\ \dot{r}_2(t) &= -d_2 r_2(t) + a_{21} h_1 \big(r_1(t) \big) + a_{22} h_2 \big(r_2(t) \big) + b_{21} h_1 \big(r_1 \big(t - \tau_1(t) \big) \big) \\ &+ b_{22} h_2 \big(r_2 \big(t - \tau_2(t) \big) \big) + k_2. \end{split}$$

The corresponding response model (3) is

$$\begin{split} \dot{s}_1(t) &= -d_1 s_1(t) + a_{11} h_1 \big(s_1(t) \big) + a_{12} h_2 \big(s_2(t) \big) + b_{11} h_1 \big(s_1 \big(t - \tau_1(t) \big) \big) \\ &+ b_{12} h_2 \big(s_2 \big(t - \tau_2(t) \big) \big) + u_1(t) + k_1, \\ \dot{s}_2(t) &= -d_2 s_2(t) + a_{21} h_1 \big(s_1(t) \big) + a_{22} h_2 \big(s_2(t) \big) + b_{21} h_1 \big(s_1 \big(t - \tau_1(t) \big) \big) \\ &+ b_{22} h_2 \big(s_2 \big(t - \tau_2(t) \big) \big) + u_2(t) + k_2. \end{split}$$

The multiplication generators are: $e_1^2 = e_2^2 = e_{12}^2 = e_1e_2e_{12} = -1$, $e_1e_2 = -e_2e_1 = e_{12}$, $e_1e_{12} = -e_2e_1 = e_{12}$. $-e_{12}e_1=-e_2, e_2e_{12}=-e_{12}e_2=e_1, r_1=r_1^0e_0+r_1^1e_1+r_1^2e_2+r_1^{12}e_{12}, r_2=r_2^0e_0+r_2^1e_1+r_2^2e_2+r_2^{12}e_{12}, r_2=r_2^0e_0+r_2^1e_1+r_2^2e_2+r_2^{12}e_1+r_2^2e_2+r_2^{12}e_1+r_2^2e_2+r_2^{12}e_1+r_2^2e_2+r_2^{12}e_1+r_2^2e_2+r_2^{12}e_1+r_2^2e_2+r_2^{12}e_1+r_2^2e_2+r_2^{12}e_1+r_2^2e_2+r_2^{12}e_1+r_2^2e_2+r_2^{12}e_1+r_2^2e_2+r_2^{12}e_1+r_2^2e_2+r_2^{12}e_1+r_2^2e_2+r_2^{12}e_1+r_2^2e_2+r_2^{12}e_1+r_2^2e_2+r_2^{12}e_1+r_2^2e_2+r_2^{12}e_1+r_2^2e_2+r_2^{12}e_1+r_2^{12}e_2+r_2^{12}e_1+r_2^{12}e_2+r_2^{12}e_2+r_2^{12}e_1+r_2^{12}e_2+r_2^{12}e_1+r_2^{12}e_2+r_2^{12}e_2+r_2^{12}e_2+r_2^{12}e_1+r_2^{12}e_2$ $s_1 = s_1^0 e_0 + s_1^1 e_1 + s_1^2 e_2 + s_1^{12} e_{12}, s_2 = s_2^0 e_0 + s_2^1 e_1 + s_2^2 e_2 + s_2^{12} e_{12}.$

Furthermore, we take

 $d_1 = 2.2$, $d_2 = 2.4$, $a_{11} = 0.5e_0 + 0.1e_1$, $a_{12} = 0.1e_0 + 0.2e_2 + 0.6e_{12},$ $a_{21} = 0.5e_0 - 0.1e_1 + 0.3e_2,$ $a_{22} = 0.3e_0 + 0.1e_1 + 0.5e_{12},$ $b_{11} = 0.1e_0 + 0.2e_1 + 0.5e_2,$ $b_{12} = 0.3e_0 + 0.1e_2 + 0.4e_{12}$ $b_{21} = 0.6e_0 - 0.2e_1 + 0.3e_2$, $b_{22} = 0.4e_0 + 0.1e_{12}$, $k_1 = 0.3e_0 + 0.1e_1 - 0.2e_{12}$ $k_2 = 0.1e_0 - 0.3e_1 + 0.1e_2 - 0.4e_{12}$

$$g_1(\boldsymbol{e}_1) = \frac{1 - e^{-\boldsymbol{e}_1^0}}{1 + e^{-\boldsymbol{e}_1^0}} e_0 + \frac{1}{1 + e^{-\boldsymbol{e}_1^1}} e_1 + \frac{1 - e^{-\boldsymbol{e}_1^2}}{1 + e^{-\boldsymbol{e}_1^2}} e_2 + \frac{1}{1 + e^{-\boldsymbol{e}_1^{12}}} e_{12},$$

$$g_2(\boldsymbol{e}_2) = \frac{1 - e^{-\boldsymbol{e}_2^0}}{1 + e^{-\boldsymbol{e}_2^0}} e_0 + \frac{1}{1 + e^{-\boldsymbol{e}_2^1}} e_1 + \frac{1 - e^{-\boldsymbol{e}_2^2}}{1 + e^{-\boldsymbol{e}_2^2}} e_2 + \frac{1}{1 + e^{-\boldsymbol{e}_2^{12}}} e_{12},$$

in which $\mathbf{e}_i(t) = s_i(t) - r_i(t)$ and $\mathbf{e}_i^A(t) = s_i^A(t) - r_i^A(t)$, i = 1, 2. The time-varying delays are considered as $\tau_1(t) = \tau_2(t) = 0.4 |\cos(t)| + 0.03$ with $\tau_1 = \tau_2 = 0.43$. Furthermore, the activation function satisfies Assumption (A1) with $l_1 = l_2 = 0.5$. By selecting $\lambda_{11} = 2.5$, $\lambda_{12} = 2.6$, $\lambda_{21} = 3.5$, $\lambda_{22} = 3.8$, $\lambda_{31} = 1.95$, $\lambda_{32} = 2$, and $\alpha = 0.5$.

Besides it is easy to obtain $d_1 = 2.2$, $d_2 = 2.4$, $a_{11}^{A,\bar{B}} = 0.6$, $a_{12}^{A,\bar{B}} = 0.9$, $a_{21}^{A,\bar{B}} = 0.7$, $a_{22}^{A,\bar{B}} = 0.9$, $b_{11}^{A,\bar{B}} = 0.8$, $b_{12}^{A,\bar{B}} = 0.8$, $b_{21}^{A,\bar{B}} = 0.7$, $b_{22}^{A,\bar{B}} = 0.5$. The initial conditions of drive-response systems (1) and (3) are taken as $\varphi_1(t) = 1.5e_0 - 1.2e_1 - 0.9e_2 + 2e_{12}$ for $t \in [-0.43, 0]$, $\varphi_2(t) = -1.6e_0 + 2.5e_1 + 2.2e_2 - 1.4e_{12}$ for $t \in [-0.43, 0]$, $\phi_1(t) = -2.5e_0 + 1.1e_1 + 2.2e_2 - 1.5e_{12}$ for $t \in [-0.43, 0]$, and $\phi_2(t) = 2.6e_0 - 2.1e_1 - 2.2e_2 + e_{12}$ for $t \in [-0.43, 0]$. By simple calculation, we have

$$\begin{split} d_1 + \lambda_{11} &- \sum_{j=1}^2 \sum_{A \in \Lambda} \sum_{B \in \Lambda} \left| a_{1j}^{A,\bar{B}} \right| l_j = 3.95 > 0, \\ d_2 + \lambda_{12} &- \sum_{j=1}^2 \sum_{A \in \Lambda} \sum_{B \in \Lambda} \left| a_{2j}^{A,\bar{B}} \right| l_j = 4.2 > 0, \\ \lambda_{31} &- \sum_{j=1}^2 \sum_{A \in \Lambda} \sum_{B \in \Lambda} \left| b_{1j}^{A,\bar{B}} \right| l_j = 1.15 > 0, \\ \lambda_{32} &- \sum_{j=1}^2 \sum_{A \in \Lambda} \sum_{B \in \Lambda} \left| b_{2j}^{A,\bar{B}} \right| l_j = 1.4 > 0, \\ \lambda_{21} &= 3.5 > 0, \lambda_{22} = 3.8 > 0. \end{split}$$

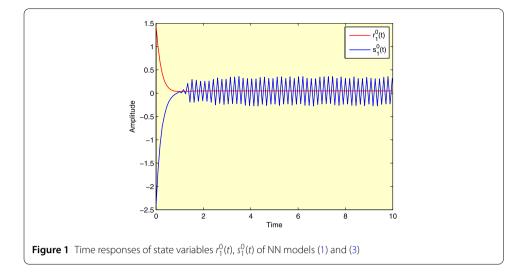
Moreover, the settling time of synchronization T^+ satisfies

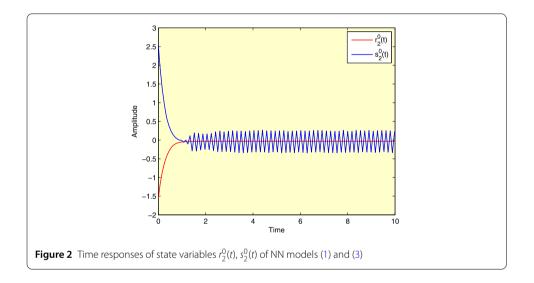
$$T^{+} = t_{0} + \frac{\mathcal{V}^{1-\alpha}(t_{0})}{\min_{i \in N}(\lambda_{2i})(1-\alpha)} = 2.1630.$$

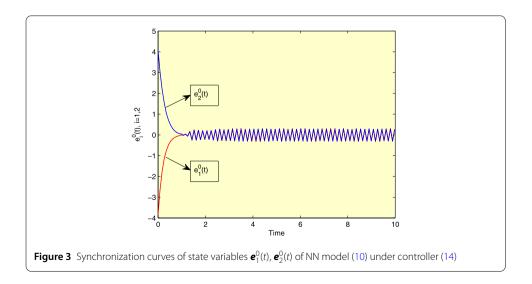
Clearly, all conditions of Theorem 3.11 are satisfied. Drive-response models (1) and (3) can achieve synchronization in finite-time with controller (14). Figures 1, 2, 4, 5, 7, 8, 10, and 11, respectively, show the time responses of the states of drive-response models (1) and (3). Besides, Figures 3, 6, 9, and 12 disclose the time responses of the states of error systems (10). From Figures 3, 6, 9, and 12, it can be seen that model (6) synchronizes with model (8) in finite-time through the controller (14) with the given initial values.

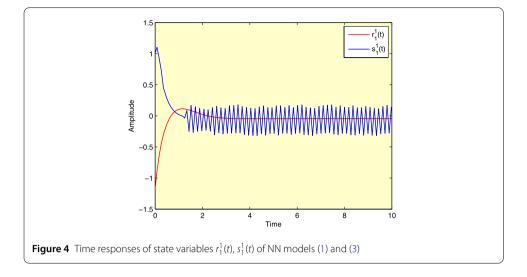
Example 2 For m = 2 and n = 2, the following two-neuron drive model (1) is considered:

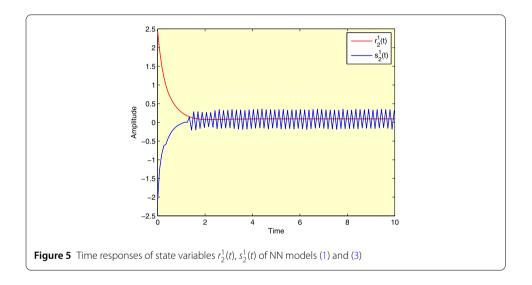
$$\begin{bmatrix} \dot{r}_1(t) \\ \dot{r}_2(t) \end{bmatrix} = -\begin{bmatrix} d_1 & 0 \\ 0 & d_2 \end{bmatrix} \begin{bmatrix} r_1(t) \\ r_2(t) \end{bmatrix} + \begin{bmatrix} a_{11} & a_{12} \\ a_{21} & a_{22} \end{bmatrix} \begin{bmatrix} h_1(r_1(t)) \\ h_2(r_2(t)) \end{bmatrix} \\ + \begin{bmatrix} b_{11} & b_{12} \\ b_{21} & b_{22}s \end{bmatrix} \begin{bmatrix} h_1(r_1(t-\tau_1(t))) \\ h_2(r_2(t-\tau_2(t))) \end{bmatrix} + \begin{bmatrix} k_1 \\ k_2 \end{bmatrix}.$$

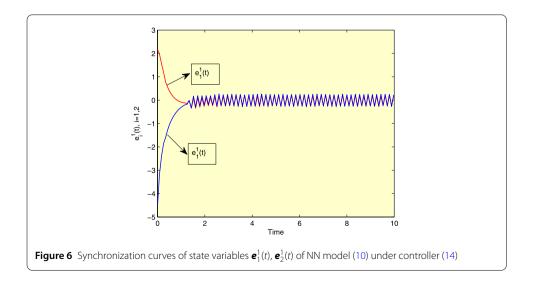


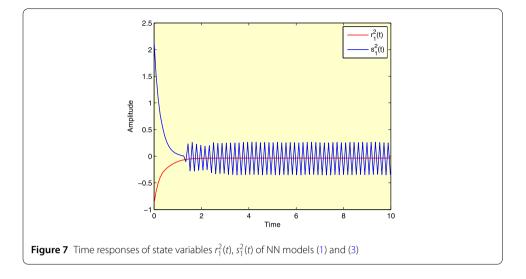


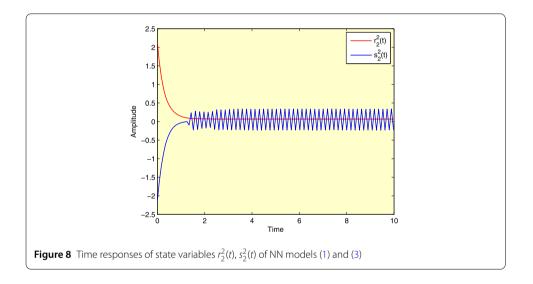


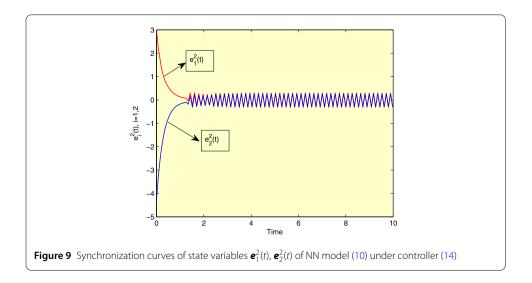


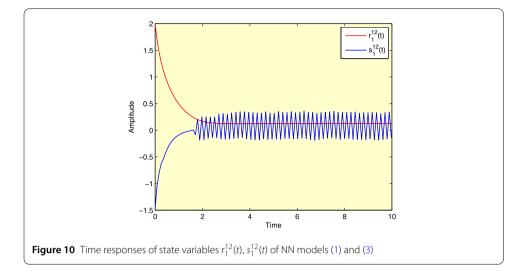


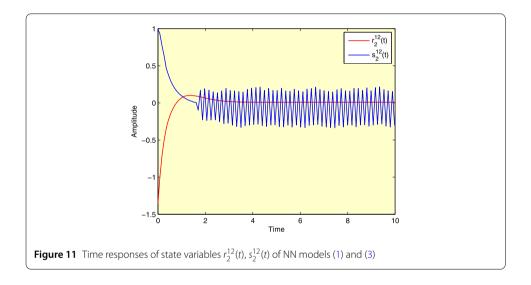


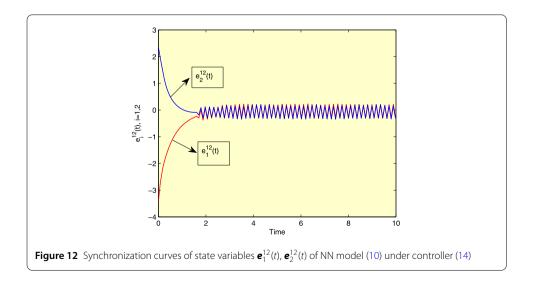












The corresponding response model (3) is

$$\begin{bmatrix} \dot{s}_{1}(t) \\ \dot{s}_{2}(t) \end{bmatrix} = - \begin{bmatrix} d_{1} & 0 \\ 0 & d_{2} \end{bmatrix} \begin{bmatrix} s_{1}(t) \\ s_{2}(t) \end{bmatrix} + \begin{bmatrix} a_{11} & a_{12} \\ a_{21} & a_{22} \end{bmatrix} \begin{bmatrix} h_{1}(s_{1}(t)) \\ h_{2}(s_{2}(t)) \end{bmatrix}$$
$$+ \begin{bmatrix} b_{11} & b_{12} \\ b_{21} & b_{22} \end{bmatrix} \begin{bmatrix} h_{1}(s_{1}(t-\tau_{1}(t))) \\ h_{2}(s_{2}(t-\tau_{2}(t))) \end{bmatrix} + \begin{bmatrix} u_{1}(t) \\ u_{2}(t) \end{bmatrix} + \begin{bmatrix} k_{1} \\ k_{2} \end{bmatrix}$$

The multiplication generators are: $e_1^2 = e_2^2 = e_{12}^2 = e_1e_2e_{12} = -1$, $e_1e_2 = -e_2e_1 = e_{12}$, $e_1e_{12} = -e_{12}e_1 = -e_2$, $e_2e_{12} = -e_{12}e_2 = e_1$, $r_1 = r_1^0e_0 + r_1^1e_1 + r_1^2e_2 + r_1^{12}e_{12}$, $r_2 = r_2^0e_0 + r_2^1e_1 + r_2^2e_2 + r_2^{12}e_{12}$, $s_1 = s_1^0e_0 + s_1^1e_1 + s_1^2e_2 + s_1^{12}e_{12}$, $s_2 = s_2^0e_0 + s_2^1e_1 + s_2^2e_2 + s_2^{12}e_{12}$.

Furthermore, we take

$$d_{1} = 2.6,$$

$$d_{2} = 2.8,$$

$$a_{11} = 0.3e_{0} - 0.1e_{1} + 0.5e_{12},$$

$$a_{12} = 0.1e_{0} + 0.3e_{1} + 0.4e_{2},$$

$$a_{21} = 0.2e_{0} + 0.6e_{2} - 0.3e_{12},$$

$$a_{22} = 0.3e_{0} + 0.5e_{12},$$

$$b_{11} = 0.4e_{0} + 0.1e_{12},$$

$$b_{12} = 0.3e_{0} - 0.1e_{2} + 0.4e_{2},$$

$$b_{21} = 0.6e_{0} + 0.2e_{1} - 0.3e_{12},$$

$$b_{22} = 0.1e_{0} + 0.2e_{1} + 0.5e_{2},$$

$$k_{1} = 0.5e_{0} - 0.2e_{1} + 0.3e_{2} - 0.1e_{12},$$

$$k_{2} = 0.4e_{0} - 0.1e_{1} + 0.3e_{12},$$

$$g_{1}(e_{1}) = \frac{1 - e^{-e_{1}^{0}}}{1 + e^{-e_{1}^{0}}}e_{0} + \frac{1}{1 + e^{-e_{1}^{1}}}e_{1} + \frac{1 - e^{-e_{1}^{2}}}{1 + e^{-e_{1}^{2}}}e_{2} + \frac{1}{1 + e^{-e_{1}^{12}}}e_{12},$$

$$g_{2}(e_{2}) = \frac{1 - e^{-e_{2}^{0}}}{1 + e^{-e_{2}^{0}}}e_{0} + \frac{1}{1 + e^{-e_{1}^{2}}}e_{1} + \frac{1 - e^{-e_{2}^{2}}}{1 + e^{-e_{2}^{2}}}e_{2} + \frac{1}{1 + e^{-e_{1}^{2}}}e_{12},$$

in which $\boldsymbol{e}_i(t) = s_i(t) - r_i(t)$ and $\boldsymbol{e}_i^A(t) = s_i^A(t) - r_i^A(t)$, i = 1, 2. By selecting $\lambda_{11} = 2.5$, $\lambda_{12} = 2.6$, $\lambda_{21} = 3.5$, $\lambda_{22} = 3.8$, $\lambda_{31} = 4.2$, $\lambda_{32} = 4.4$, $\lambda_{41} = 1.95$, $\lambda_{42} = 2$, $\alpha = 0.5$, and $\beta = 1.5$

The time-varying delays are considered as $\tau_1(t) = \tau_2(t) = 0.5|\cos(t)| + 0.02$ with $\tau_1 = \tau_2 = 0.52$. Furthermore, the activation function satisfies Assumption (A1) with $l_1 = l_2 = 0.5$. Besides it is easy to obtain $d_1 = 2.6$, $d_2 = 2.8$, $a_{11}^{A,\bar{B}} = 0.7$, $a_{12}^{A,\bar{B}} = 0.8$, $a_{21}^{A,\bar{B}} = 0.5$, $a_{22}^{A,\bar{B}} = 0.8$, $b_{11}^{A,\bar{B}} = 0.5$, $b_{21}^{A,\bar{B}} = 0.5$, $b_{21}^{A,\bar{B}} = 0.5$, $b_{22}^{A,\bar{B}} = 0.8$. The initial conditions of drive-response models (1) and (3) are taken as $\varphi_1(t) = -1.5e_0 + 1.8e_1 - 0.9e_2 + 2e_{12}$ for $t \in [-0.52, 0]$, $\varphi_2(t) = 1.6e_0 - 2e_1 + 2.2e_2 - 1.4e_{12}$ for $t \in [-0.52, 0]$, $\phi_1(t) = -2.5e_0 + 1.7e_1 + 2.2e_2 - 1.5e_{12}$ for $t \in [-0.52, 0]$, and $\phi_2(t) = 2.6e_0 + 1.2e_1 - 2.2e_2 + e_{12}$ for $t \in [-0.52, 0]$. By simple computation, we have

$$\begin{aligned} d_1 + \lambda_{11} - \sum_{j=1}^2 \sum_{A \in \Lambda} \sum_{B \in \Lambda} \left| a_{1j}^{A,\bar{B}} \right| l_j &= 4.35 > 0 \\ d_2 + \lambda_{12} - \sum_{j=1}^2 \sum_{A \in \Lambda} \sum_{B \in \Lambda} \left| a_{2j}^{A,\bar{B}} \right| l_j &= 4.75 > 0 \\ \lambda_{41} - \sum_{j=1}^2 \sum_{A \in \Lambda} \sum_{B \in \Lambda} \left| b_{1j}^{A,\bar{B}} \right| l_j &= 1.4 > 0, \\ \lambda_{42} - \sum_{j=1}^2 \sum_{A \in \Lambda} \sum_{B \in \Lambda} \left| b_{2j}^{A,\bar{B}} \right| l_j &= 1.35 > 0, \\ \lambda_{21} &= 3.5 > 0, \qquad \lambda_{22} &= 3.8 > 0, \\ \lambda_{31} &= 4.2 > 0, \qquad \lambda_{32} &= 4.4 > 0. \end{aligned}$$

Clearly, all conditions of Theorem 3.12 are satisfied. Therefore, drive-response models (6) and (8) achieve fixed-time synchronization with controller (24). Moreover, the settling time of synchronization T^+_{max} satisfies

$$T_{\max}^{+} = \frac{1}{\Pi_2(1-\alpha)} + \frac{1}{\Pi_1(\beta-1)} = 0.2291.$$

5 Conclusion

In this article, we have studied the finite-/fixed-time synchronization for Clifford-valued RNN models with time-varying delays. In order to overcome the difficulty of the noncommutativity of the multiplication of Clifford numbers, we first decomposed the considered Clifford-valued drive and response models into real-valued drive and response models. Besides, suitable time-delayed feedback controllers have been constructed to examine the synchronization problem associated with the finite-/fixed-time error models. By utilizing the finite-/fixed-time stability concepts, some computational techniques, new synchronization criteria have been derived through appropriate Lyapunov functions to guarantee that the drive-response models achieve synchronization in finite-/fixed-time. Finally, we have also presented numerical examples to illustrate the effectiveness of the results. The results obtained in this paper can be further extended to other complex systems. We would like to extend our results to more general Clifford-valued NN models, such as Cohen-Grossberg Clifford-valued NNs, Clifford-valued inertial NNs, Clifford-valued high-order Hopfield NNs, and fuzzy Clifford-valued NNs. Moreover, we will focus on the problem of global stabilization analysis of Clifford-valued NN models with the help of various control systems. The corresponding results will be carried out in the near future.

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Availability of data and materials

Data sharing is not applicable to this article as no datasets were generated or analysed during the current study.

Competing interests

The authors declare that they have no competing interests.

Authors' contributions

Funding acquisition, NB; Conceptualization, GR; Software, GR and NB; Formal analysis, GR and NB; Methodology, GR and NB; Supervision, RS, PA, GR and CPL; Writing–original draft, GR; Validation, GR; Writing–review and editing, GR and NB. All authors have read and agreed to the published version of the manuscript.

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